

# \*Sheet 1 (1)

① using partial fraction technique find  $e(0)$  and  $e(10)$ , then check the value of  $e(0)$  using the initial value theorem. For:-

$$E(z) = \frac{1}{(z-1)(z-0.3)}$$

$$E(z) = \frac{A z z^{-1}}{z-1} + \frac{B z z^{-1}}{z-0.3} \longrightarrow e(k) = A u(k-1) + B (0.3)^{k-1} u(k-1)$$

$$e(k) = u(k-1) \left[ \frac{10}{7} - \frac{10}{7} (0.3)^{k-1} \right]$$

$$\therefore e(0) = u(-1) \left[ \frac{10}{7} - \frac{10}{7} (0.3)^{-1} \right] = 0$$

$$\therefore e(10) = u(9) \left[ \frac{10}{7} - \frac{10}{7} (0.3)^9 \right] = 1.428$$

$$\therefore e(0) = \lim_{z \rightarrow \infty} E(z) = \lim_{z \rightarrow \infty} \frac{1}{(z-1)(z-0.3)} = 0$$

② A function  $e(t) = A \cos(\omega t)$  is sampled every  $T=0.2$  sec. If the Z-transform of the resultant number sequence is find  $A$  and  $\omega$  for:-

$$E(z) = \frac{3z(z-0.6967)}{z^2-1.3934z+1}$$

$$\cos(\omega t) = \frac{z(z - \cos \omega T)}{z^2 - 2z \cos \omega T + 1} \quad \therefore A = 3$$

$$\cos(\omega T) = 0.6967 \longrightarrow \omega T = [\cos^{-1} 0.6967] \times \frac{\pi}{180} = 0.8$$

$$\omega = \frac{0.8}{0.2} = 4 \text{ rad/sec}$$

③ Solve the given difference equation for  $y(k)$  using:-  
the sequential technique and the Z-transform for:-

$$y(k+2) - \frac{3}{4} y(k+1) + \frac{1}{8} y(k) = e(k) \quad , y(0) = y(1) = 0 \quad e(k) = 1, k=0, 1, 2, \dots$$

$$z^2 y(z) - \underbrace{z^2 y(0)}_{\rightarrow 0} - \underbrace{z y(1)}_{\rightarrow 0} - \frac{3}{4} z y(z) + \frac{3}{4} \underbrace{z y(0)}_{\rightarrow 0} + \frac{1}{8} y(z) = E(z)$$

$$y(z) \left[ z^2 - \frac{3}{4} z + \frac{1}{8} \right] = \frac{z}{z-1}$$

$$\therefore y(z) = z \frac{1}{(z-1)(z-0.5)(z-0.25)} = \frac{z A}{(z-1)} + \frac{z B}{z-0.5} + \frac{z C}{z-0.25}$$

$$y(z) = \frac{8/3 z}{z-1} + \frac{8 z}{z-0.5} + \frac{16/3 z}{z-0.25}$$

$$y(k) = 8/3 u(k) - 8 (0.5)^k u(k) + \frac{16}{3} (0.25)^k u(k)$$



$$y(0) = \lim_{z \rightarrow 1} (z-1) f(z) = \lim_{z \rightarrow 1} (z-1) \frac{z}{(z-1)(z-0.5)(z-0.25)} = \frac{1}{(0.5)(0.75)} = \frac{8}{3}$$

4] A function  $e(t)$  sampled, and the resultant o/p sequence has the following Z-transform, find Z.T of  $e(t-3T)u(t-3T)$

, find Z.T of  $e(t+T)u(t+T)$

$$E(z) = \frac{z^3}{z^3 + 3z^2 + 5z - 9}$$

$$e(t-3T)u(t-3T) = \frac{z^{-3} z^3}{z^3 + 3z^2 + 5z - 9}$$

$$Z[e(t+T)u(t+T)] = \frac{z^4}{z^3 + 3z^2 + 5z - 9} - z e(0)$$

$$e(0) = \lim_{z \rightarrow 0} \frac{1}{1 - \frac{3}{z} + \frac{5}{z^2} - \frac{9}{z^3}} = 1$$

5] Given a discrete time dynamic system represented by the difference equation solve for  $x(k)$  as a function of time  $k$

$$x(k+2) + 3x(k+1) + 2x(k) = e(k)$$

$$e(k) = \begin{cases} 1 & , k=0 \\ 0 & , \text{otherwise} \end{cases}$$

$$\text{with } x(0) = 0, x(1) = -1$$

$$z^2 x(z) + z^2 x(0) - z x(1) + 3z x(z) + 2x(z) = E(z)$$

$$z^2 x(z) + z + 3z x(z) + 2x(z) = 1$$

$$x(z) = \frac{1-z}{z^2 + 3z + 2} = \frac{A}{z+1} + \frac{B}{z+2} = \frac{2z z^{-1}}{z+1} - \frac{3z z^{-1}}{z+2}$$

↓ Z.T

$$x(k) = 2(-1)^{k-1} u(k-1) - 3(-2)^{k-1} u(k-1)$$



$$\frac{1}{s} \rightarrow 1 \rightarrow \frac{z}{z-1}$$

$$\frac{1}{s^2} \rightarrow t \rightarrow \frac{Tz}{(z-1)^2}$$

$$\frac{1}{s+1} \rightarrow e^{-T} \rightarrow \frac{z}{z-e^{-T}}$$

$$\frac{1}{s+2} \rightarrow e^{-2T} \rightarrow \frac{z}{z-e^{-2T}}$$

\* Sheet 2 (1)

Find the T.F  $\frac{C(z)}{E(z)}$   $E(s) \rightarrow \downarrow \begin{matrix} E^*(s) \\ \downarrow \end{matrix} \begin{matrix} \boxed{G_1(s)} \\ \downarrow \end{matrix} \begin{matrix} \boxed{\frac{1}{(s+1)(s+2)}} \\ \downarrow \end{matrix} C(s)$

$$\frac{C(z)}{E(z)} = Z \left[ \frac{1-e^{-Ts}}{s(s+1)(s+2)} \right] = 1-z^{-1} Z \left[ \frac{1}{s(s+1)(s+2)} \right] = [1-z^{-1}] Z \left[ \frac{0.5}{s} + \frac{1}{s+1} + \frac{0.5}{s+2} \right]$$

$$= \frac{z-1}{z} \left[ \frac{0.5z}{z-1} - \frac{z}{z-e^{-T}} + \frac{0.5z}{z-e^{-2T}} \right] = 0.5 - \frac{z-1}{(z-e^{-T})} + \frac{0.5(z-1)}{(z-e^{-2T})}$$

[2] if  $D(z) = \frac{z}{z-1}$  find T.F  $\rightarrow D(s) = \frac{1}{s}$

$$\frac{C(z)}{E(z)} = Z \left[ \frac{1-e^{-Ts}}{s^2(s+1)} \right] = \frac{z-1}{z} Z \left[ \frac{A}{s} + \frac{B}{s^2} + \frac{C}{s+1} \right]$$

$$= \frac{z-1}{z} \left[ \frac{A z}{z-1} + \frac{B z}{(z-1)^2} + \frac{C z}{z-e^{-T}} \right]$$

[3] if  $G_1(s) = G_2(s) = \frac{1-e^{-Ts}}{s(s+1)}$  find  $\frac{C(z)}{E(z)}$   $E(s) \rightarrow \downarrow \begin{matrix} \boxed{G_1(s)} \\ \downarrow \end{matrix} \begin{matrix} \boxed{G_2(s)} \\ \downarrow \end{matrix} C(s)$

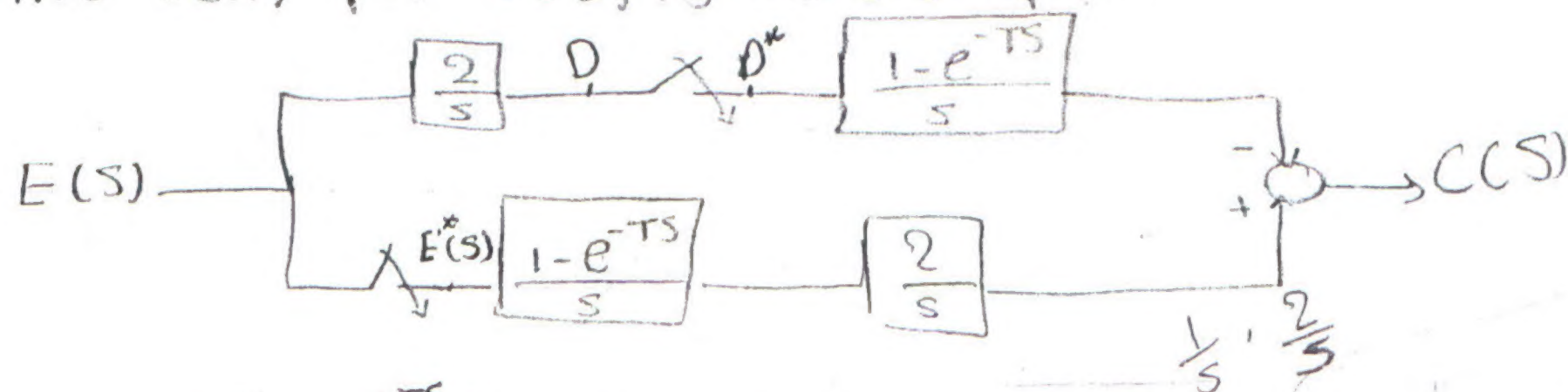
$$\frac{C(z)}{E(z)} = G_1(z) \cdot G_2(z) = Z \left[ \frac{1-e^{-Ts}}{s(s+1)} \right] \cdot Z \left[ \frac{1-e^{-Ts}}{s(s+1)} \right]$$

$$= \frac{z-1}{z} Z \left[ \frac{1}{s} - \frac{1}{s+1} \right] \cdot \frac{z-1}{z} Z \left[ \frac{1}{s} - \frac{1}{s+1} \right]$$

$$= \frac{z-1}{z} \left[ \frac{z}{z-1} - \frac{z}{z-e^{-T}} \right] \cdot \frac{z-1}{z} \left[ \frac{z}{z-1} - \frac{z}{z-e^{-T}} \right]$$

$$= \left[ 1 - \frac{z-1}{z-e^{-T}} \right]^2$$

[4] Find CCR for C(t) is unit step

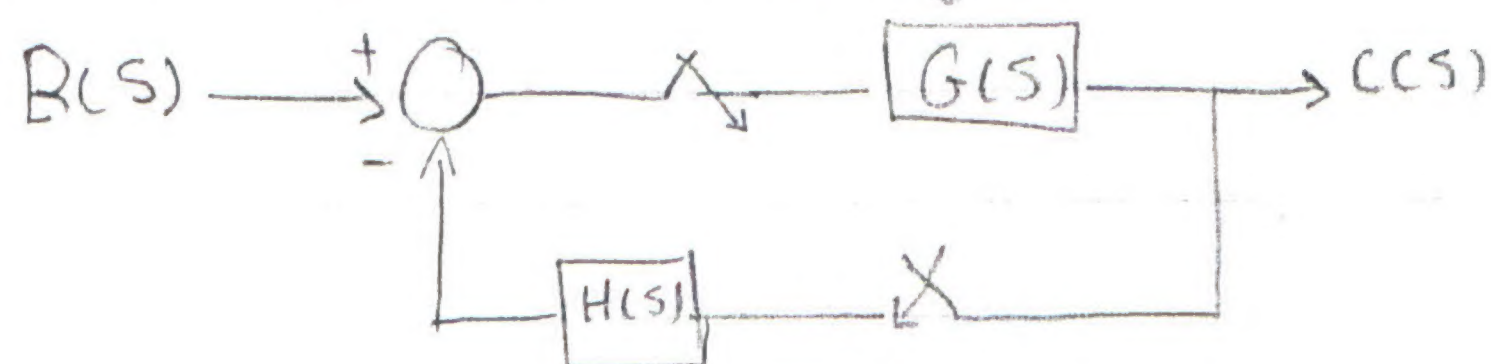


$$C(z) = Z \left[ \frac{(1-e^{-Ts})^2}{s^2} \right] Z[E(s)] - Z \left[ E(s) \frac{2}{s} \right] \cdot Z \left[ \frac{1-e^{-Ts}}{s} \right]$$

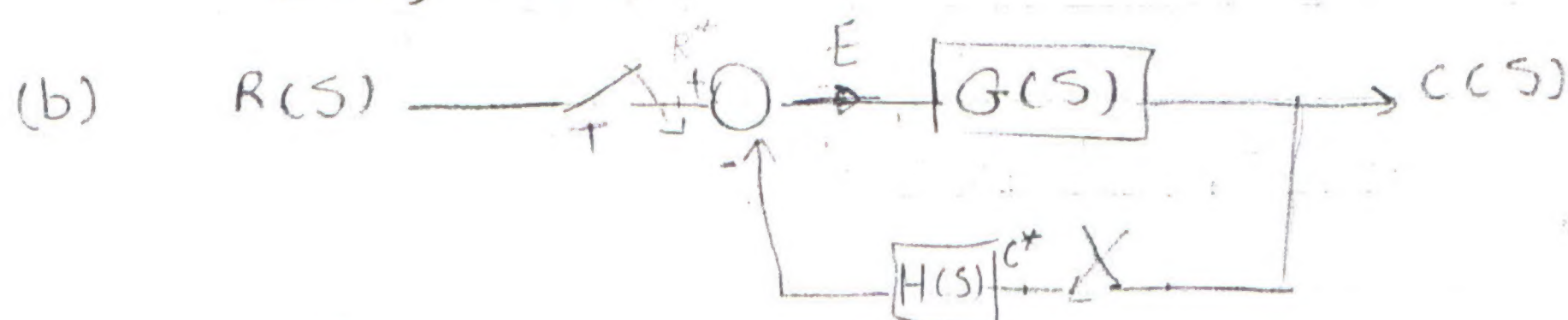
$$= 2 \frac{z-1}{z} \frac{z}{(z-1)^2} \frac{z}{z-1} - 2 \frac{z(z)}{(z-1)^2} \cdot \frac{z-1}{z} \cdot \frac{z}{z-1} = \frac{2z-2z^2}{(z-1)^2}$$



5] (a) Find the  $C(Z)$  for the system



$$C(Z) = \frac{G(Z)}{1 + G(Z) \cdot H(Z)} \cdot R(Z)$$



$$C(s) = G(s) [R^*(s) - H(s) \cdot C^*(s)]$$

$$(1 + H) C^*(s) = G^*(s) R^*(s)$$

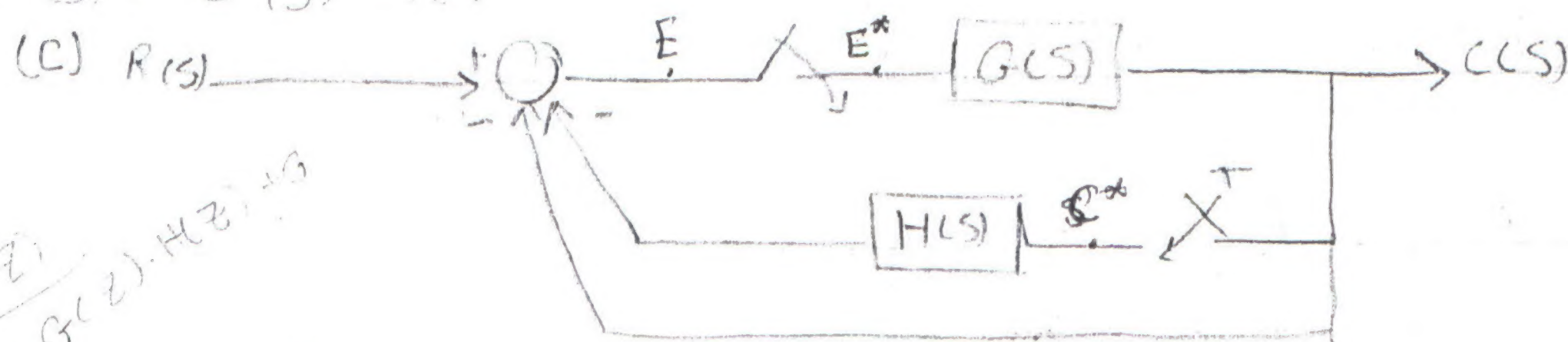
$$C(s) = G(s) (R^*(s) - H^* C^*(s))$$

$$C^*(s) = \frac{G^*(s)}{1 + H^*(s)} R^*(s)$$

$$C^*(s) = G^*(s) (R^*(s) - H^* C^*(s))$$

$$E(s) = E(s) G(s)$$

$$C^*(s) = E^*(s) G^*(s)$$



$$\frac{G(Z)}{1 + G(Z) \cdot H(Z)}$$

$$C(s) = E^*(s) \cdot G^*(s)$$

$$E(s) = R(s) - H(s) \cdot C^*(s) - C(s)$$

$$E(s) = R(s) - H(s) \cdot C^*(s) - E^*(s) \cdot G(s)$$

Starting

$$C^*(s) = E^*(s) \cdot G^*(s)$$

$$\rightarrow E^*(s) = \frac{C^*(s)}{G^*(s)}$$

$$E^*(s) = R^*(s) - H^*(s) \cdot C^*(s) - E^*(s) \cdot G^*(s)$$

$$E^*(s) [1 + G^*(s)] = R^*(s) - H^*(s) \cdot C^*(s)$$

$$C^*(s) [1 + G^*(s)] = R^*(s) \cdot G^*(s) - H^*(s) \cdot C^*(s) \cdot G^*(s)$$

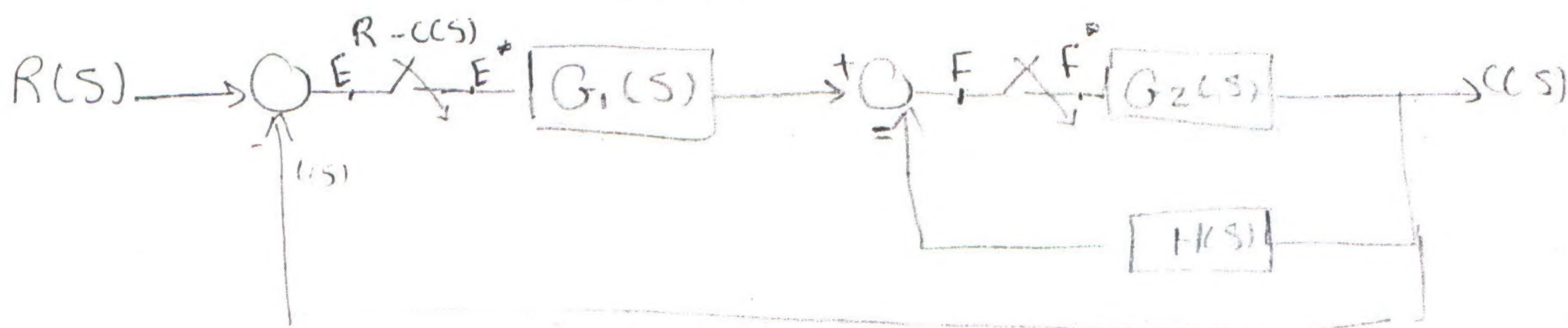
$$C^*(s) = \frac{G^*(s) \cdot R^*(s)}{1 + G^*(s) + G^*(s) \cdot H^*(s)}$$

$$1 + G^*(s) + G^*(s) \cdot H^*(s)$$



#sheet 2 (2)

Q Find the overall T.f =  $\frac{C(z)}{R(z)}$



$$C(s) = F^*(s) \cdot G_2(s)$$

$$E(s) = R(s) - C(s)$$

$$F(s) = G_1(s) \cdot E^*(s) - C(s) \cdot H(s)$$

$$\therefore E(s) = R(s) - F^*(s) \cdot G_2(s)$$

$$F(s) = G_1(s) \cdot E^*(s) - F^*(s) \cdot H(s) \cdot G_2(s)$$

↓ Starling

$$C^*(s) = F^*(s) \cdot G_2^*(s)$$

$$\longrightarrow F^*(s) = \frac{C^*(s)}{G_2^*(s)}$$

$$E^*(s) = R^*(s) - F^*(s) \cdot G_2^*(s)$$

$$F^*(s) = G_1^*(s) \cdot E^*(s) - F^*(s) \cdot H G_2^*(s)$$

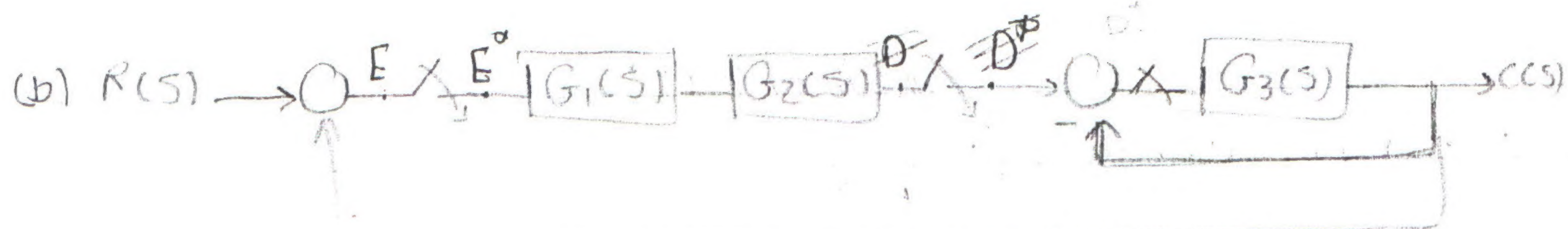
$$F^*(s) [1 + H G_2^*(s)] = G_1^*(s) \cdot E^*(s)$$

$$\frac{F^*(s)}{F^*(s)} [1 + H G_2^*(s)] = \frac{G_1^*(s) R^*(s) - G_1^*(s) G_2^*(s) \cdot F^*(s)}{F^*(s)}$$

$$F^*(s) [1 + H G_2^*(s) + G_1^*(s) \cdot G_2^*(s) \cdot F^*(s)] = G_1^*(s) \cdot R^*(s)$$

$$C^*(s) [1 + H G_2^*(s) + G_1^* G_2^* F^*(s)] = G_1^* G_2^* R^*(s)$$





$$C(s) = D^*(s) \cdot G_3(s)$$

$$D(s) = G_1(s) \cdot G_2(s) \cdot E^*(s) - C(s)$$

$$E(s) = R(s) - C(s)$$

$$D(s) = G_1 G_2(s) E^*(s) - D^*(s) \cdot G_3(s)$$

$$E(s) = R(s) - D^*(s) \cdot G_3(s)$$

Starling

$$C^*(s) = D^*(s) \cdot G_3^*(s)$$

$$D^*(s) = C^*(s) / G_3^*(s)$$

$$D^*(s) = G_1 G_2(s) E^*(s) - D^*(s) \cdot G_3^*(s)$$

$$E^*(s) = R^*(s) - D^*(s) \cdot G_3^*(s)$$

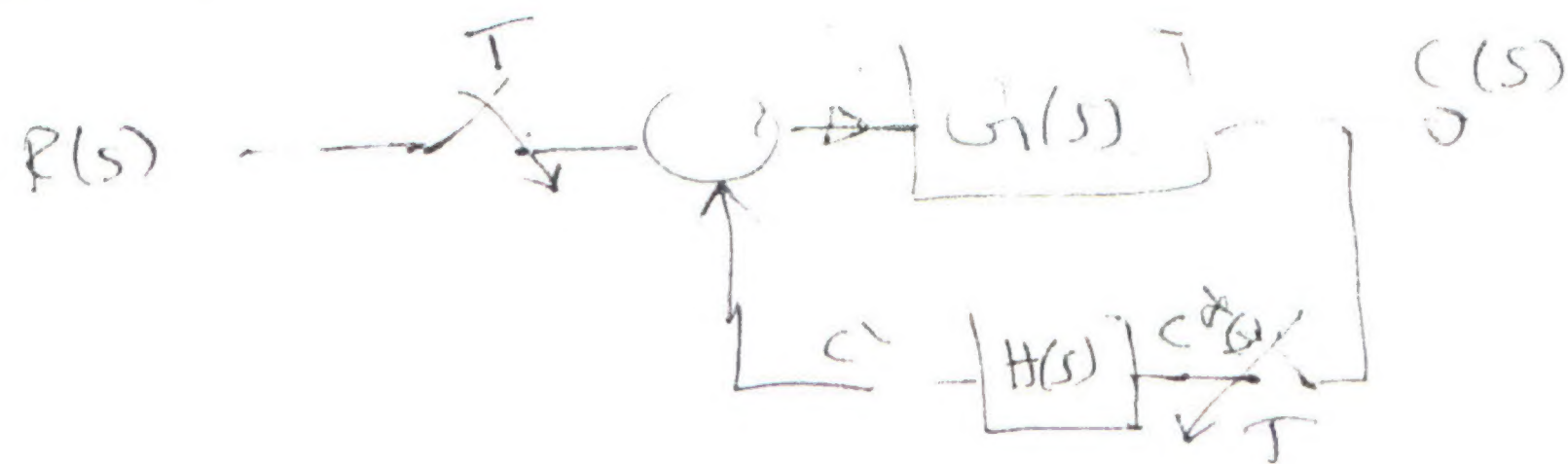
$$D^*(s) [1 + G_3^*(s)] = G_1 G_2(s) E^*(s)$$

$$D^*(s) [1 + G_3^*(s)] = G_1 G_2(s) R^*(s) - G_1 G_2(s) G_3^*(s) \cdot D^*(s)$$

$$D^*(s) [1 + G_3^*(s) + G_1 G_2(s) \cdot G_3^*(s)] = G_1 G_2(s) R^*(s)$$

$$C^*(s) [1 + G_3^*(s) + G_1 G_2(s) G_3^*(s)] = G_1 G_2(s) \cdot G_3^*(s) R^*(s)$$

5. b



$$C(s) = G(s) [R^*(s) - E^*(s) H(s)]$$

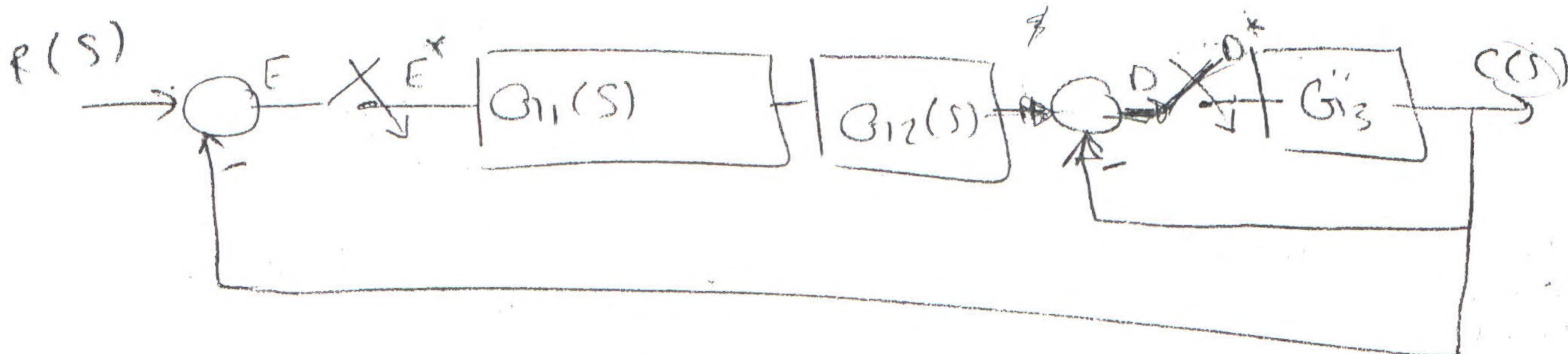
$$C(s) = G(s) R^*(s) - E^*(s) H(s) G(s)$$

$$E^*(s) = G^*(s) R^*(s) - C^*(s) \overline{GH}^*(s)$$

$$(1 + \overline{GH}^*(s)) C^*(s) = G^*(s) R^*(s)$$

$$\frac{C^*(s)}{R^*(s)} = \frac{G^*(s)}{1 + \overline{GH}^*(s)}$$





$$C(s) = G_3 D^*$$

$$D = G_1(s) G_2(s) E^* - C(s)$$

$$E = R(s) - C(s)$$

$$C^*(s) = G_3^* D^* \rightarrow$$

$$D^* = \overline{G_1 G_2}^* E^* - C^*(s)$$

$$E^* = R^* - C^*(s)$$

$$E^* = R^* - G_3^* D^*$$

$$E^* = R^* - \overline{G_1 G_2 G_3}^* E^* - G_3^* C^*(s)$$

$$[1 + \overline{G_1 G_2 G_3}]^* E^* = R^* - G_3^* C^*(s)$$

$$E^* = R^* - G_3^* C^*$$

$$C^*(s) + G_3^*(s) C^* =$$

$$\overline{G_1 G_2 G_3}^* R^* -$$

$$\overline{G_1 G_2 G_3}^* C^*$$

$$D^* + C^*(s) = \overline{G_1 G_2}^* (R^* - C^*)$$

$$\frac{C^*(s)}{G_3^*(s)} + C(s) = \overline{G_1 G_2}^* (R^* - C^*)$$

$$C^* \left( \frac{1 + G_3^*(s)}{\overline{G_1 G_2 G_3}^*(s)} \right) = \overline{G_1 G_2 G_3}^*$$

$$\frac{C(z)}{R(z)} = \frac{G_1 G_2(z) G_3(z)}{1 + G_1 G_2(z) G_3(z) + G_3(z)}$$

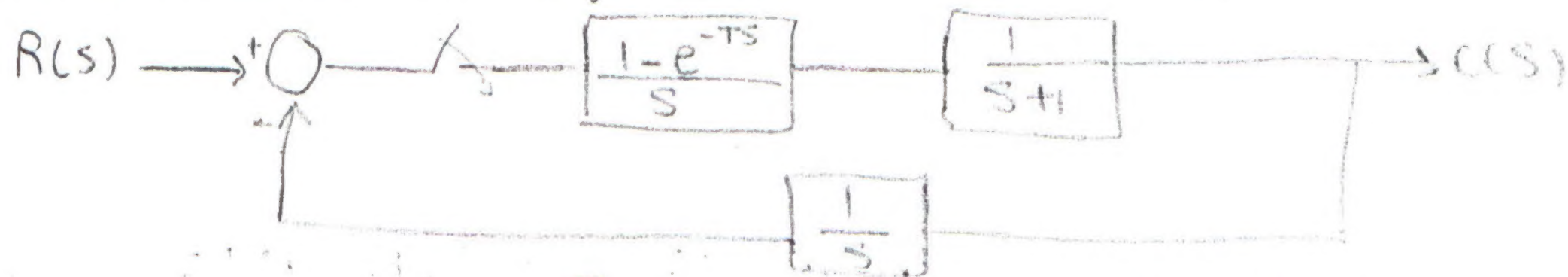
$$R = \frac{wz}{1}$$

$$f(z)$$



# # Sheet (3) (1)

□ Determine the output for the unit step I/P



$$\begin{aligned} \text{T.F} = \frac{C(z)}{R(z)} &= \frac{Z \left[ \frac{1-e^{-Ts}}{s(s+1)} \right]}{1 + Z \left[ \frac{1-e^{-Ts}}{s^2(s+1)} \right]} = \frac{(1-z^{-1})Z \left[ \frac{A}{s} + \frac{B}{s+1} \right]}{1 + (1-z^{-1})Z \left[ \frac{A}{s} + \frac{B}{s^2} + \frac{C}{s+1} \right]} \\ &= \frac{(1-z^{-1})Z \left[ \frac{1}{s} + \frac{1}{s+1} \right]}{1 + (1-z^{-1})Z \left[ \frac{1}{s} + \frac{1}{s+1} + \frac{1}{s^2} \right]} = \frac{\frac{z-1}{z} \left[ \frac{z}{z-1} - \frac{z}{z-e^{-1}} \right]}{1 + \frac{z-1}{z} \left[ \frac{z}{z-1} + \frac{z}{z-e^{-1}} + \frac{z}{(z-1)^2} \right]} \\ &= \frac{1 - \frac{z-1}{z-e^{-1}}}{1 + \left[ 1 + \frac{z-1}{z-e^{-1}} + \frac{1}{z-1} \right]} = \frac{\frac{z-e^{-1}}{z-e^{-1}}}{1 + \left[ \frac{(z-e^{-1})(z-1) + (z-1)^2 + (z-e^{-1})}{(z-e^{-1})(z-1)} \right]} \\ &= \frac{\frac{1-e^{-1}}{z-e^{-1}}}{\frac{(z-e^{-1})(z-1) + (z-e^{-1})(z-1) + (z-1)^2 + (z-e^{-1})}{(z-e^{-1})(z-1)}} = \frac{(1-e^{-1})(z-1)}{(z-e^{-1})(z-1) + (z-e^{-1})(z-1) + (z-1)^2 + (z-e^{-1})} \end{aligned}$$

$$C(z) = \frac{0.632(z-1)}{z^2 - z + 0.632} R(z)$$

$$C(z) = \frac{0.632(z-1)}{z^2 - z + 0.632} \cdot \frac{z}{z-1} = \frac{0.632 z}{z^2 - z + 0.632} \quad [\cos \sin w]$$

$$\sin(wt) \xrightarrow{Z.T} \frac{z \sin wT}{z^2 - 2z \cos wT + 1}$$

$$a^T \sin(wt) \xrightarrow{Z.T} \frac{z/a \sin wT}{(z/a)^2 - 2(z/a) \cos wT + 1} = \frac{a \sin w \cdot z}{z^2 - 2a \cos w \cdot z + a^2}$$

$$a^2 = 0.632 \rightarrow a = 0.796$$

$$2a \cos w = 1 \rightarrow w = 61.027 = 0.89 \text{ rad}$$

$$a \sin w = 0.618$$

$$C(z) =$$



$$CH(z) = 1 + 0.1.T.F$$

$$GH(z)$$

$$= GH(z)$$

2) Determine the Steady-State error when A

1- unit step I/P

$$O.L.T.F = \overline{GH(z)} = Z \left[ \frac{1-e^{-Ts}}{s^2(s+1)} \right] = \frac{Z-1}{Z} Z \left[ \frac{A}{s} + \frac{B}{s^2} + \frac{C}{s+1} \right]$$

$$\frac{Z-1}{Z} Z \left[ \frac{1}{s^2} + \frac{1}{s+1} - \frac{1}{s} \right] = \frac{Z-1}{Z} \left[ \frac{Z}{Z-0.367} + \frac{Z}{(Z-1)^2} - \frac{Z}{Z-1} \right]$$

$$= \frac{Z-1}{Z-0.367} + \frac{1}{Z-1} - 1 = \frac{(Z-1)^2 + (Z-0.367) - (Z-1)(Z-0.367)}{(Z-1)(Z-0.367)}$$

$$\overline{GH(z)} = \frac{Z^2 - 2Z + 1 + Z - 0.367 - Z^2 + 1.367Z + 0.367}{(Z-1)(Z-0.367)} = \frac{0.367Z + 1}{(Z-1)(Z-0.367)}$$

$$e_{ss} \mid_{\text{unit step}} = \lim_{Z \rightarrow 1} \frac{1}{1+Kp} = 0 = 0$$

$$Kv = \lim_{Z \rightarrow 1} (Z-1) \overline{GH(z)} = \frac{1.367}{1-0.367} \quad \checkmark \quad e_{ss} = \frac{1}{Kv}$$

Steady State o/p for unit step I/P

$$T.F = \frac{C(z)}{R(z)} = \frac{GH(z)}{1+GH(z)} \quad \checkmark$$

$$C(z) = \checkmark \propto \frac{Z}{Z-1}$$

$$C_{ss} = \lim_{Z \rightarrow 1} (Z-1) C(z) = \frac{(Z-e^{-1})(Z-1) + (Z-1)^2 + (Z-e^{-1})}{(Z-e^{-1})(Z-1)}$$

$$\frac{Z^2 - Z + 0.632 + Z^2 + 2Z - 1 + Z - 0.367}{Z^2 - Z + 0.632} = \frac{(Z-0.367)(Z-1)}{Z^2 - 1.367Z + 0.367}$$

$$Z^2 - 1.367Z + 0.367 \quad (Z^2 + 2Z - 1 + Z - 0.367)$$

$$\frac{1.633Z - 1}{Z^2 - 1.367Z + 0.367}$$



sheet (3) (2)

$$\boxed{3} \quad \frac{C(z)}{R(z)} = \frac{0.4}{s(s+1)}$$

@ impulse response  $\longrightarrow R(z) = 1$

$$C(z) = \frac{0.4}{s(s+1)} \quad R(z) = \frac{A}{z} + \frac{B}{s+1} = 0.4 z \left[ \frac{A}{s} + \frac{B}{s+1} \right] = 0.4 z \left[ \frac{1}{s} - \frac{1}{s+1} \right]$$

$$= 0.4 \left[ \frac{z}{z-1} - \frac{z}{z-0.367} \right] = \frac{0.4 z (z-0.367) - 0.4 z (z-1)}{(z-1)(z-0.367)}$$

$$C(k) = 0.4 u(k) - 0.4 (0.367)^{k-1}$$

@ Discuss the stability  $|z| < 1$

$$= \lim_{z \rightarrow 1} (z-1) \frac{0.4 (z-0.367) - 0.4 z (z-1)}{(z-1)(z-0.367)} = \frac{0.2532}{0.633} = 0.4$$

critical stable.

$$\boxed{4} \quad @ \text{O.L.T.F} = \frac{C(z)}{R(z)} = \frac{z-1}{z} z \left[ \frac{k}{s^2(s+1)} \right] = k \frac{0.367 z + 1}{(z-1)(z-0.367)}$$



# Sheet 4 (1)

Check if the roots of the following char eq lie within the unit circle.

a)  $5z^2 - 2z + 3 = 0$

$(z - (0.2 + 0.75i))(z - (0.2 - 0.75i)) = 0$

$|z_{1,2}| = \sqrt{(0.2)^2 + (0.75)^2} = 0.776 < 1$

The system is stable  $\rightarrow$  lie on unit circle.

b)  $z^3 - 0.2z^2 - 0.25z + 0.05 = 0$

$(z + 0.5)(z - 0.5)(z - 0.2) = 0$

$|z_1| = 0.5 \quad |z_2| = 0.5 \quad |z_3| = 0.2$

all poles located in unit circle

the system is stable.

c)  $z^4 - 1.7z^3 + 1.04z^2 - 0.268z + 0.024 = 0$

~~$z^4 - 1.7z^3$~~  using Jury Test

(1)  $f(1) \rightarrow 1 - 1.7 + 1.04 - 0.268 + 0.024 = 0.096 > 0$

(2)  $(-1)^4 f(-1) \rightarrow 1 + 1.7 + 1.04 + 0.268 + 0.024 = 4.032 > 0$

(3)  $|a_0| = 0.024 < |a_n| = 1$

(4) Jury Test

$b_0 = \begin{vmatrix} 0.024 & 1 \\ 1 & 0.024 \end{vmatrix} = -0.99$

$b_1 = \begin{vmatrix} 0.024 & -1.7 \\ 1 & -0.268 \end{vmatrix} = 1.693$

$b_2 = \begin{vmatrix} 0.024 & 1.04 \\ 1 & 1.04 \end{vmatrix} = -1.05$

$b_3 = \begin{vmatrix} 0.024 & -0.268 \\ 1 & -1.7 \end{vmatrix} = 0.227$

$c_0 = \begin{vmatrix} -0.99 & 0.227 \\ 0.227 & -0.99 \end{vmatrix} = 0.946$

$c_2 = \begin{vmatrix} -0.99 & 1.693 \\ 0.227 & -1.05 \end{vmatrix} = 0.629$

	$z^0$	$z^1$	$z^2$	$z^3$	$z^4$
1	0.024	-0.268	1.04	-1.7	1
2	1	-1.7	1.04	-0.268	0.024
3	$b_0$	$b_1$	$b_2$	$b_3$	
4	$b_3$	$b_2$	$b_1$	$b_0$	
5	$c_0$	$c_1$	$c_2$		

(4)  $|b_0| = 0.99 > |b_3| = 0.227$

$|c_0| = 0.946 > |c_2| = 0.629$

The system is stable  
all roots lie on unit circle



$$(d) z^3 + 5z^2 + 3z + 2 = 0$$

$$(z + 4.42)(z - (-0.29 + 0.61i))(z + (0.29 + 0.61i)) = 0$$

$$|z_1| = 4.42 < 1$$

$$|z_{2,3}| = \sqrt{(0.29)^2 + (0.61)^2} = 0.675 < 1$$

the pole lie on unit Circle

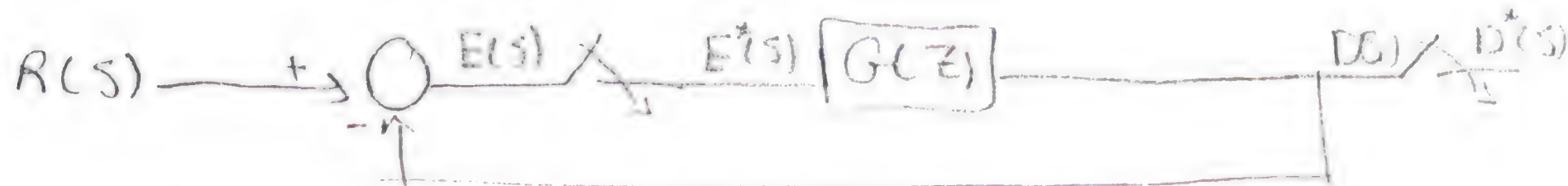
the system is stable.



sheet 4(2)

Q For the system shown with  $G(z) = \frac{0.1(z+0.9)}{(z-1)(z-0.7)}$   
Check the stability using Jury, bilinear.

$$T.f = \frac{G(z)}{1+G(z)}$$



$$\text{Ch. eqn} = 1 + GH(z) = 0$$

$$(z-1)(z-0.7) + 0.1z + 0.09 = 0$$

$$z^2 - 1.7z + 0.7 + 0.1z + 0.09 = 0$$

$$z^2 - 1.6z + 0.79 = 0$$

1) Bilinear  $z = \frac{1+r}{1-r}$

$$\left(\frac{1+r}{1-r}\right)^2 - 1.6\left(\frac{1+r}{1-r}\right) + 0.79 = 0 \quad | \cdot (1-r)^2$$

$$(1+r)^2 - 1.6(1-r^2) + 0.79(1-r)^2 = 0$$

$$1+2r+r^2 - 1.6 + 1.6r^2 + 0.79 - 1.58r + 0.79r^2 = 0$$

$$3.39r^2 + 1.58r + 0.19 = 0$$

$$r^2 \quad | \quad 3.39 \quad 0.19$$

$$r \quad | \quad 1.58 \quad 0$$

$$r^0 \quad | \quad 0.19$$

→ stable

2) Jury Test

$$\text{Ch. eqn} \rightarrow z^2 - 1.6z + 0.79 = 0$$

1)  $f(1) = 1 - 1.6 + 0.79 = 0.19 > 0 \quad \checkmark$

2)  $(-1)^2 f(-1) = 1 + 1.6 + 0.79 = 3.39 > 0 \quad \checkmark$

3)  $|a_0| = 0.79 < |a_2| = 1 \quad \checkmark$

the system is stable.



$$\frac{1}{s} + \frac{B}{s+2} = \frac{A}{s} + \frac{B}{s+2}$$

$$B=0.5 \quad \text{at } s=0 \quad 0 + 5 = -1 \quad -A + \frac{1}{2} + \frac{1}{2} = -1 \quad A = -0.25$$

3) write the char eqn for the system, check stability



$$O.L.T.F = 10Z \left[ \frac{1-e^{-Ts}}{s^2(s+2)} \right] = 10(1-Z^{-1})Z \left[ \frac{A^{1/2}}{s} + \frac{B^{1/2}}{s^2} + \frac{C^{1/4}}{s+2} \right] \frac{1}{s^2(s+2)}$$

$$= 10(1-Z^{-1})Z[Au(t) + B e^{+t} + C e^{-2t}]$$

$$= 10 \left[ \frac{Z-1}{Z} \right] \left[ \frac{A Z}{Z-1} + \frac{B Z}{(Z-1)^2} + \frac{C Z}{Z-e^{-2}} \right]$$

$$= 10 \left[ A^{1/4} + \frac{B^{1/2}}{(Z-1)} + \frac{C(Z-1)}{(Z-e^{-2})} \right] = -5/2 + \frac{5}{Z-1} + \frac{5}{2} \frac{(Z-1)}{(Z-e^{-2})}$$

$$= \frac{2.5(Z-1)(Z-0.135) + 5(Z-0.135) + 2.5(Z-1)^2}{(Z-1)(Z-0.135)} \quad Z^2 - 2Z + 1$$

$$= \frac{-2.5Z^2 + 2.83Z - 0.3375 + 5Z - 0.675 + 2.5Z^2 - 5Z + 2.5}{Z^2 + 1.135Z + 0.135}$$

$$= \frac{2.83Z + 2.2}{(Z-1)(Z-0.135)}$$

$$\text{Char eqn} = 1 + O.L.T.F = (Z-1)(Z-0.135) + 2.83Z + 2.2 = 0$$

$$= Z^2 - 1.135Z + 0.135 + 2.83Z + 2.2$$

$$f(Z) = Z^2 + 1.7Z + 2.335$$

Jury Test

$$① f(1) = 1 + 1.7 + 2.335 > 0 \quad \checkmark$$

$$② f(1)^2 f(-1) = 1 - 1.7 + 2.335 > 0 \quad \checkmark$$

$$③ |a_0| = 2.335 < |a_n| = 1 \quad \checkmark$$

the system is stable.



$$\text{Chleqn} = 1 + 0.1 \text{ T.F}$$

HGH(z) sheet 4(3)

4) for the following system, determine the range of K for stability

(a)

$$\frac{C(s)}{R(s)} = 0.1 \text{ T.F} = \overline{GH}(z) = Z \left[ \frac{(1 - e^{-Ts})K}{s^2(s+2)} \right] = (1 - z^{-1})Z \left[ \frac{K}{s^2(s+2)} \right]$$

$$= K (1 - z^{-1})Z \left[ \frac{A}{s} + \frac{B}{s^2} + \frac{C}{s+2} \right] = K(1 - z^{-1})Z \left[ \frac{0.25}{s} + \frac{0.5}{s^2} + \frac{0.25}{s+2} \right]$$

$$= K(1 - z^{-1}) \left[ \frac{0.25Z}{z-1} + \frac{0.5Z}{(z-1)^2} + \frac{0.25Z}{Z - e^{-2}} \right]$$

$$= K(1 - z^{-1}) \left[ 0.25 + \frac{0.5}{(z-1)} + \frac{0.25(z-1)}{Z - e^{-2}} \right]$$

$$= K \frac{-0.25(z-1)(z - e^{-2}) + 0.5(z - e^{-2}) + 0.25(z-1)^2}{(z-1)(z - e^{-2})} \quad e^{-2} = 0.135$$

$$= K \frac{-0.25(z^2 - z - e^{-2}z + e^2) + 0.5z - 0.5e^{-2} + 0.25z^2 - 0.5z + 0.25}{(z-1)(z - e^{-2})}$$

$$= K \frac{-0.25z^2 + 0.25z + 0.0337z - 0.0337 + 0.6z - 0.067 + 0.25z^2 - 0.5z + 0.25}{(z-1)(z - e^{-2})}$$

$$= K \frac{0.2837z + 0.2167}{(z-1)(z - 0.135)}$$

$$\text{Chleqn} \rightarrow 1 + 0.1 \text{ T.F} = 1 + \frac{K(0.2837z + 0.2167)}{(z-1)(z - 0.135)} = 0$$

$$(z-1)(z - 0.135) + K(0.2837z + 0.2167)$$

$$z^2 - 1.135z + 0.135 + 0.2837Kz + 0.2167K = 0$$

$$z^2 + z[0.2837K - 1.135] + 0.135 + 0.2167K = 0$$

$$① f(1) > 0 \rightarrow 1 + 0.2837K - 1.135 + 0.135 + 0.2167K > 0$$

$$0.5004K > 0 \quad K > 0$$

$$② (-1)^2 f(-1) > 0 \rightarrow 1 - 0.2837K + 1.135 + 0.135 + 0.2167K > 0$$

$$2.27 - 0.067K > 0 \rightarrow 0.067K < 2.27$$

$$(K < 33.881)$$

$$③ |a_0| = 1 < |a_m| = 0.135 + 0.2167K$$

$$0.135 + 0.2167K > 1 \quad K > 3.9927$$





$$(b) \text{ O.L.T.F.} = Z \left[ \frac{(1-e^{-Ts})K(e^{-Ts})}{s^2(s+1)} \right] \cdot K(1-z^{-1}) Z \left[ \frac{e^{-Ts} - e^{-2Ts}}{s^2(s+1)} \right]$$

$$1 - e^{-Ts} = 1 - z^{-1}$$

$$K(1-z^{-1})z^{-1}$$

$$Z \left[ \frac{1}{s^2(s+1)} \right]$$



\*Sheet(5)(1)

III Calculate the  $e_{ss}$  for A unit Step, unit Ramp i/p for o.l.t.f.

(a)  $G H(z) = \frac{0.0952 K z}{(z-1)(z-0.965)}$  Type 1

$$e_{ss} = \lim_{z \rightarrow 1} (z-1) \frac{R(z)}{1+G H(z)}$$

1- for unit step i/p  $\rightarrow \frac{1}{K_p}$

$$K_p = \lim_{z \rightarrow 1} G H(z) = \lim_{z \rightarrow 1} \frac{0.0952 K z}{(z-1)(z-0.965)} = \infty$$

$$\therefore e_{ss} = 1 + \frac{1}{\infty} = 0 \quad e_{ss} = \frac{1}{\infty} = 0$$

2- for unit ramp i/p  $\rightarrow \frac{1}{K_v}$

$$K_v = \frac{1}{T} \lim_{z \rightarrow 1} (z-1) G H(z) = \lim_{z \rightarrow 1} \frac{0.0952 K z}{z-0.965} = \frac{0.0952 K}{0.035} = 2.72 K$$

$$\therefore e_{ss} = \frac{1}{2.72 K}$$

3- for  $r(t) = 3u(t) + 2t \rightarrow R(z) = \frac{3z}{z-1} + 2 \frac{Tz}{(z-1)^2}$

$$\begin{aligned} e_{ss} \Big|_{r(t)=3u(t)+2t} &= 3 \left( \frac{1}{1+K_p} \right) + 2 \left( \frac{1}{K_v} \right) \\ &= 3 + 0 + 2 \frac{1}{2.72 K} = \frac{25}{34 K} \end{aligned}$$



$$(b) G H(z) = \frac{z+0.9}{(z)(z-0.5)} \rightarrow \text{Type (0)}$$

1. For unit step input  $\rightarrow e_{ss} = \frac{1}{1+K_p}$   
 $K_p = \lim_{z \rightarrow 1} \frac{z+0.9}{z(z-0.5)} = \frac{1.9}{0.5} = 3.8$

$$e_{ss} = \frac{1}{1+3.8} = \frac{1}{4.8} = 0.208$$

2. For unit ramp input  $\rightarrow e_{ss} = \frac{1}{K_v}$   
 $K_v = \frac{1}{T} \lim_{z \rightarrow 1} (z-1) \frac{(z+0.9)}{z(z-0.5)} = 0$

$$e_{ss} = \frac{1}{0} = \infty$$

3. For  $r(t) = 3u(t) + 2t$

$$e_{ss} \Big|_{r(t)=3u(t)+2t} = 3 \left( \frac{1}{1+K_p} \right) + 2 \left( \frac{1}{K_v} \right) = 3 + 0.208 + 2 \times \infty = \infty$$

$$(c) G H(z) = \frac{(z+1)}{(1-z)^2} = \frac{z+1}{(-(z-1))^2} = \frac{z+1}{(z-1)^2} \rightarrow \text{Type (2)}$$

1. For unit step  $\rightarrow e_{ss} = \frac{1}{1+K_p}$   
 $K_p = \lim_{z \rightarrow 1} \frac{z+1}{(z-1)^2} = \infty$

$$e_{ss} = \frac{1}{1+\infty} = 0$$

2. For unit ramp input  $\rightarrow e_{ss} = \frac{1}{K_v}$   
 $K_v = \frac{1}{T} \lim_{z \rightarrow 1} (z-1) \frac{(z+1)}{(z-1)^2} = \frac{2}{0} = \infty$

$$e_{ss} = \frac{1}{\infty} = 0$$

3. For  $e_{ss} \Big|_{r(t)=3u(t)+2t}$

$$= 3 \left( \frac{1}{1+K_p} \right) + 2 \left( \frac{1}{K_v} \right) = 0$$



sheet (5) (2)

$$a) G H(z) = \frac{z}{(z^2 - 1)(z^2 - z + 0.5)}$$

$$= \frac{z}{(z-1)(z+1)(0.5z+0.5i)(0.5z-0.5i)} \rightarrow \text{Type 1}$$

1- for unit step i/p  $\rightarrow e_{ss} = \frac{1}{1+K_P}$

$$K_P = \lim_{z \rightarrow 1} G H(z) = \frac{1}{0} = \infty$$

$$\therefore e_{ss} = \frac{1}{1+\infty} = 0$$

2- for unit ramp i/p  $\rightarrow e_{ss} = \frac{1}{K_V}$

$$K_V = \lim_{z \rightarrow 1} (z-1) G H(z) = \frac{1}{(2)(0.5)} = 1$$

$$e_{ss} = \frac{1}{1} = 1$$

3- for  $r(t) = 3u(t) + 2t$

$$e_{ss}|_{r(t)=3u(t)+2t} = 3 \frac{1}{K_P + 1} + 2 \frac{1}{K_V} = 2$$

2] Draw root locus & Calculate range of  $K$  for stability for the following op

$$ii) G H(z) = \frac{0.0952 K z}{(z-1)(z-0.905)}$$

1)  $n_p = 2 \rightarrow 1, 0.905$

$n_z = 1 \rightarrow 0$

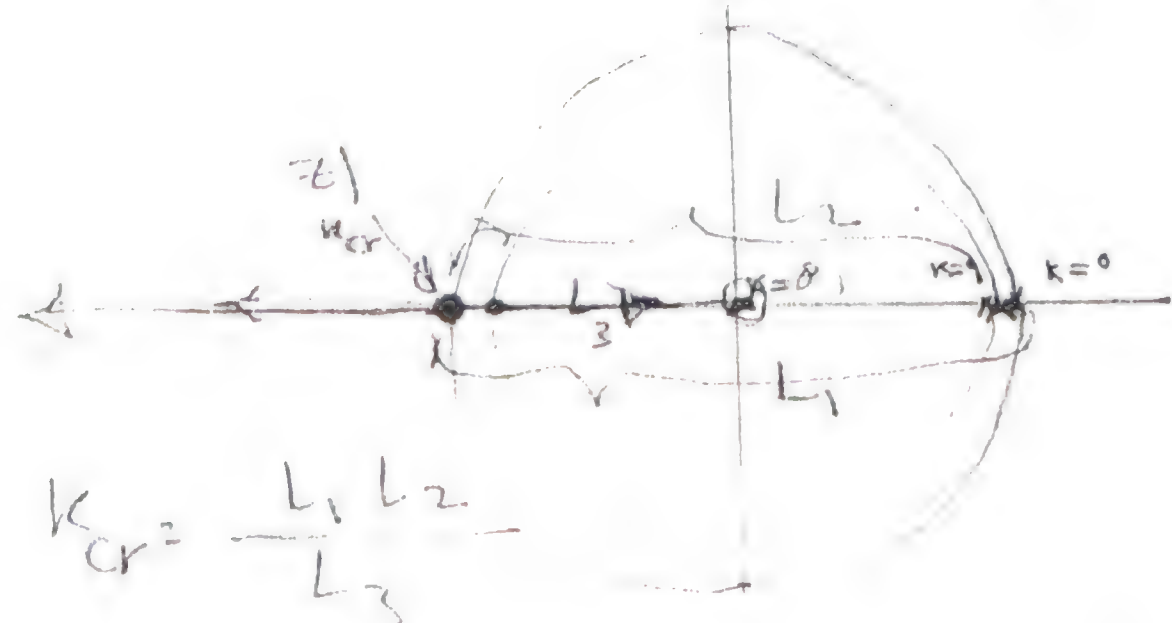
2) real part 1)  $0.905 \rightarrow 1$   
2)  $0 \rightarrow -\infty$

3) Asymptotic lines

$$N_0 = n_p - n_z = 1$$

$$\sigma_0 = \frac{\sum \text{poles} - \sum \text{zeros}}{1} = 1 + 0.905 - 0 = 1.905$$

$\phi = 180^\circ$



$$K_{cr} = \frac{L_1 L_2}{L_3}$$

$$K = f(z)$$





#### 4) Break points

Ch eqn  $\rightarrow 1 + G H(z) = 0$

$$1 + \frac{0.0952Kz}{(z-1)(z-0.905)} = 0 \rightarrow (z-1)(z-0.905) + 0.0952Kz = 0$$

$$z^2 - 1.905z + 0.905 + 0.0952Kz = 0$$

$$K = - \frac{(z^2 - 1.905z + 0.905)}{0.0952z} = - [10.5z - 20.01 + 9.51z^{-1}]$$

$$\frac{dK}{dz} = 0 \rightarrow 10.5 - \frac{9.51}{z^2} = 0 \rightarrow \frac{10.5z^2 - 9.51}{z^2} = 0$$

$$z^2 = \frac{9.51}{10.5} \rightarrow z = \pm 0.952, -0.952$$

$$\therefore r = \frac{0.952 + 0.952}{2} = 0.952, \text{ Center} = 0$$

system stable.

$\rightarrow$  Determining Kcr using Jury

$$\text{ch eqn} \rightarrow \underset{a_2=1}{z^2} + \underset{a_1}{(0.952K - 1.905)}z + \underset{a_0}{0.905} = 0$$

[1]  $F(1) > 0$

$$1 + 0.952K - 1.905 + 0.905 > 0$$

$$|K > 0|$$

[2]  $(-1)^2 F(-1) > 0$

$$-0.952K + 1.905 + 0.905 > 0 \rightarrow -0.952K + 2.81 > 0$$

$$K < \frac{2.81}{0.952} \quad |K < 2.951|$$

[3]  $|a_0| < |a_n|$

$$0.905 < 1$$

$$0 < K < 2.951$$



sheet(5)(3)

$$GH(z) = \frac{k(z+0.9)}{(z-1)(z-0.7)}$$

$$① n_p = 2 \rightarrow 1, 0.7$$

$$n_z = 1 \rightarrow -0.9$$

$$2) \text{ real part } ① 0.7 \rightarrow 1$$

$$② -0.9 \rightarrow -\infty$$

③ Asymptotic lines

$$N_0 = 1$$

$$f_0 = \frac{1 \cdot 0.7 + 0.9}{1} = 2.6$$

$$Q = 180^\circ$$

④ Break points

$$\text{Characteristic} \rightarrow 1 + GH(z) = 0$$

$$1 + k GH(z) = 0$$

$$(z-1)(z-0.7) + k(z+0.9) = 0$$

$$k GH(z) = -1 \quad k = \frac{-1}{GH(z)}$$

$$z^2 - 1.7z + 0.7 + k(z+0.9) = 0$$

$$k = \frac{z^2 - 1.7z + 0.7}{z - 0.9} \Rightarrow \frac{dk}{dz} = \frac{(z-0.9)(2z-1.7) - (z^2 - 1.7z + 0.7)(1)}{(z-0.9)^2} = 0$$

$$2z^2 - 3.5z + 1.53 - z^2 + 1.7z - 0.7 = 0$$

$$z^2 - 1.8z + 0.83 = 0 \quad z_{1,2} = 0.9 \pm 0.14i$$

$$k = \frac{-1(z-1)(z-0.7)}{(z+0.9)}$$

$$\frac{dk}{dz} = 0$$

$$\frac{dk}{dz} = \frac{(z+0.9)(2z-1.7) - (z^2 - 1.7z + 0.7)(1)}{(z+0.9)^2} = 0$$

$$(z+0.9)(2z-1.7) = z^2 - 1.7z + 0.7$$

$$z =$$

$$0.9 \pm 0.14i$$





$$5) G_H(z) = \frac{0.15K(z+0.7453)}{z(z-1)(z-0.4119)}$$

$$1) n_p = 3 \quad 0, 1, 0.4119$$

$$n_z = 1 \quad -0.7453$$

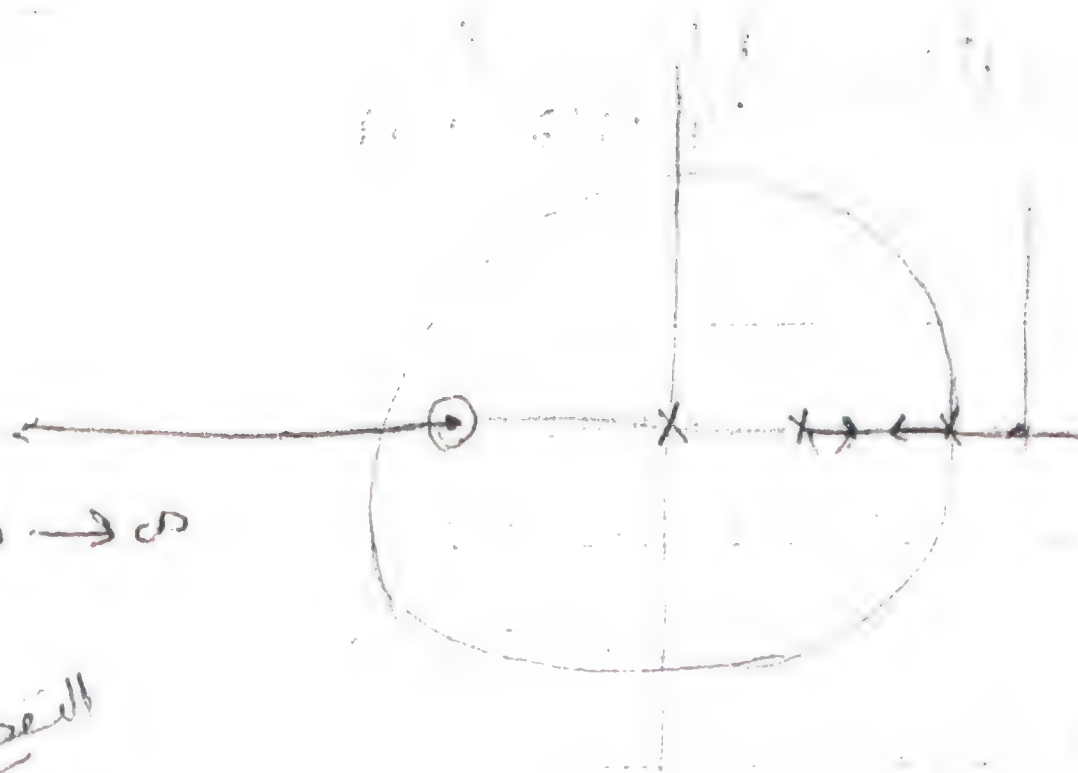
$$2) \text{Real Part } \textcircled{1} 1 \rightarrow 0.4119 \quad \textcircled{2} -0.7453 \rightarrow \infty$$

3) asymptotic lines

$$\textcircled{1} N_0 = 3 - 1 = 2$$

$$\sigma_0 = \frac{0 + 1 + 0.4119 + 0.7453}{2} = 1.0786$$

$$\phi_0 = -90^\circ, 90^\circ$$



4) Break point  $1 + G_H(z) = 0$

$$z(z-1)(z-0.4119) + 0.15Kz + 0.111K = 0$$

$$z(z^2 - 1.4119z + 0.4119) + 0.15Kz + 0.111K = 0$$

$$z^3 - 1.4119z^2 + 0.4119z + 0.15Kz + 0.111K = 0$$

$$z^3 - 1.4119z^2 + 0.5619z + 0.111K = 0$$

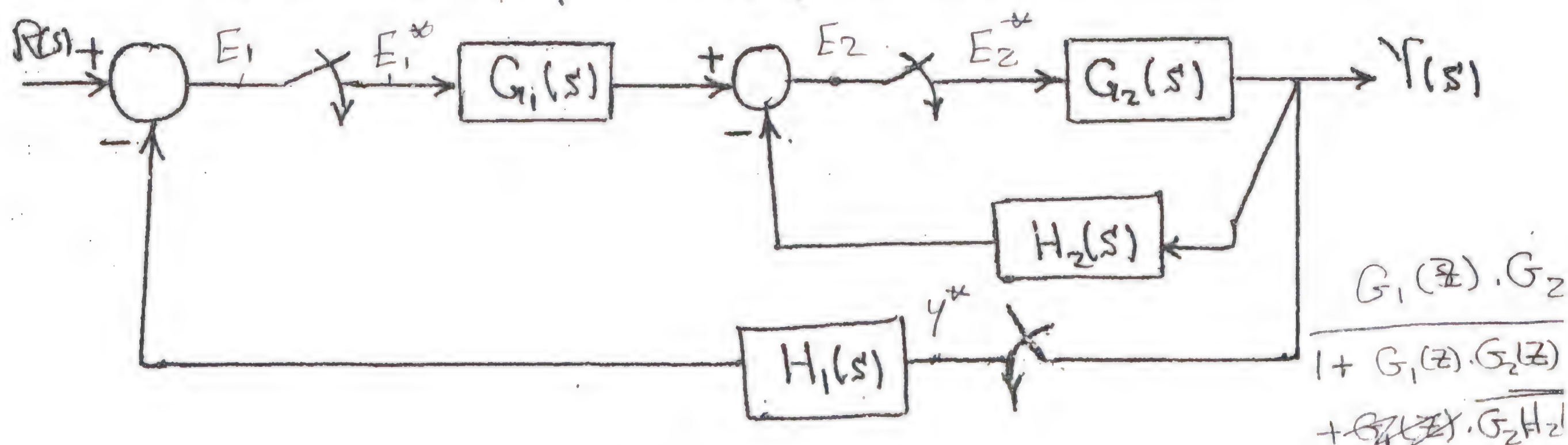
$$K = -\frac{z(z-1)(z-0.4119)}{(z+0.7453)} = \frac{z^3 - 1.4119z^2 + 0.4119z}{z+0.7453}$$

$$\frac{dK}{dz} = \frac{(z+0.7453)(3z^2 - 2.8238z + 0.4119) - (z^3 - 1.4119z^2 + 0.4119z)}{(z+0.7453)^2}$$

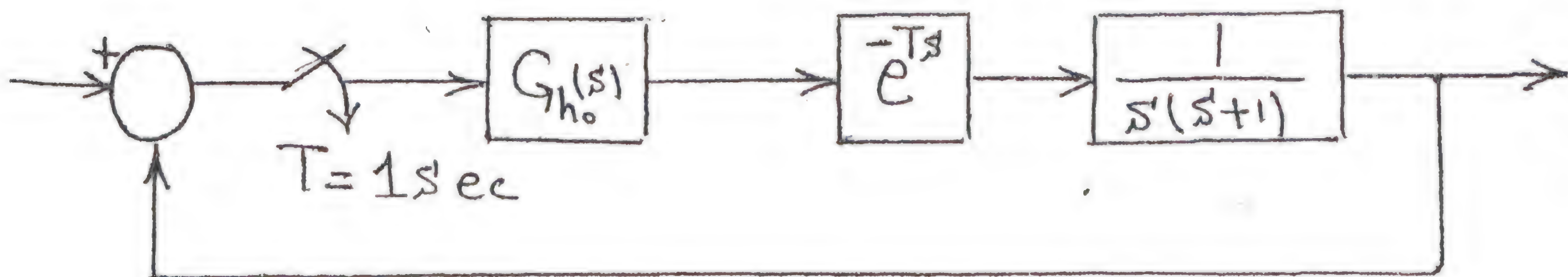


mid term

- 1] For the following block diagram, find, if it exists, the closed loop transfer function.



- 2] For the following unity feedback control systems,



- (1) Determine the open loop transfer function.
- (2) What is the type of the system and calculate the error constants.
- (3) Find the steady state error for unit step i/p.
- (4) Determine the closed loop T.F.
- (5) Calculate the steady state value of the output for unit step i/p.
- (6) Check the system stability.



Sheet (8) (1)

II] Draw the Bode Diagram of the following functions, Determine  $G_M, P_M$ .

(a)  $G H(z) = \frac{0.5(z+0.1)}{(z-0.7)(z-0.9)} \rightarrow \text{put } z = \frac{1+r}{1-r}$

$$G H(r) = \frac{0.5\left(\frac{1+r}{1-r} + 0.1\right)}{\left(\frac{1+r}{1-r} - 0.7\right)\left(\frac{1+r}{1-r} - 0.9\right)} = 0.5 \frac{\frac{1+r+0.1-0.1r}{1-r}}{\frac{1+r-0.7+0.7r}{1-r} \cdot \frac{1+r-0.9+0.9r}{1-r}}$$

$$= 0.5 \left[ \frac{(1.1+0.9r)(1-r)}{(0.3+1.7r)(0.1+1.9r)} \right] = \frac{0.5 \times 0.1}{0.3 \times 0.1} \left[ \frac{1 + \frac{r}{1.22}}{\left(1 + \frac{r}{0.18}\right)\left(1 + \frac{r}{0.05}\right)} \right]$$

$$G H(\omega) = 18.33 \left[ \frac{\left(1 + \frac{j\omega}{1.22}\right)(1-j\omega)}{\left(1 + \frac{j\omega}{0.18}\right)\left(1 + \frac{j\omega}{0.05}\right)} \right]$$

Term	$\phi(\omega r)$	
18.33	0	
$1 + j \frac{\omega r}{1.22}$	$\tan^{-1} \frac{\omega r}{1.22}$	
$1 - j\omega$	$-\tan^{-1}(\omega r)$	
$\frac{1}{1 + j \frac{\omega r}{0.18}}$	$-\tan^{-1}\left(\frac{\omega r}{0.18}\right)$	
$\frac{1}{1 + j \frac{\omega r}{0.05}}$	$-\tan^{-1}\left(\frac{\omega r}{0.05}\right)$	

$$\phi(\omega r) = \tan^{-1}\left(\frac{\omega r}{1.22}\right) - \tan^{-1}(\omega r) - \tan^{-1}\left(\frac{\omega r}{0.18}\right) - \tan^{-1}\left(\frac{\omega r}{0.05}\right)$$

$\omega r$	0	1.22	1	0.18	0.05	$\infty$
$\phi$	0	-174.919	-172.593	-121.288	-61.039	-180



$$(b) G_H(z) = \frac{0.5(z-1)}{(z-0.1)(z-0.8)} \rightarrow \text{put } z = \frac{1+r}{1-r}$$

$$G_H(r) = \frac{0.5 \left( \frac{1+r}{1-r} - 1 \right)}{\left( \frac{1+r}{1-r} - 0.1 \right) \left( \frac{1+r}{1-r} - 0.8 \right)} = \frac{0.5(1+r-1+r)/1-r}{\frac{(1+r-0.1+0.1r)(1+r-0.8+0.8r)}{(1-r)^2}}$$

$$= \frac{0.5(2r)(1-r)}{(0.9+0.1r)(0.2+0.8r)} = 5.55 \frac{r(1-r)}{\left(1 + \frac{r}{0.818}\right) \left(1 + \frac{r}{0.111}\right)}$$

$$G_H(j\omega) = 5.55 \frac{j\omega(1-j\omega)}{\left(1 + j\frac{\omega}{0.818}\right) \left(1 + j\frac{\omega}{0.111}\right)}$$

Term	$\phi(\omega r)$	
5.55	0	$\} 20 \log 5.55$
$j\omega r$	$+90$	$\} +20 \text{ dB/dec}$
$1+j\omega r$	$\tan^{-1} \omega r$	$\} +20 \text{ dB/dec}$
$1/(1+j\frac{\omega r}{0.818})$	$-\tan^{-1} \frac{\omega r}{0.818}$	$\} -20 \text{ dB/dec}$
$1/(1+j\frac{\omega r}{0.111})$	$-\tan^{-1} \frac{\omega r}{0.111}$	$\} -20 \text{ dB/dec}$

$$Q(\omega r) = 0 + 90 + \tan^{-1}(\omega r) - \tan^{-1} \frac{\omega r}{0.818} - \tan^{-1} \frac{\omega r}{0.111}$$

31.8199    115.211

							-65.53	
$\omega r$	0	1	0.818	0.111	0.5	1.5	2	$\infty$
$Q(\omega r)$	90	0.617	20.108	47.505	7.646	-0.8527	-1.143	



sheet (b) (2)

$$(c) G H(z) = \frac{0.5(z + 0.76)}{(z-1)(z-0.45)}$$



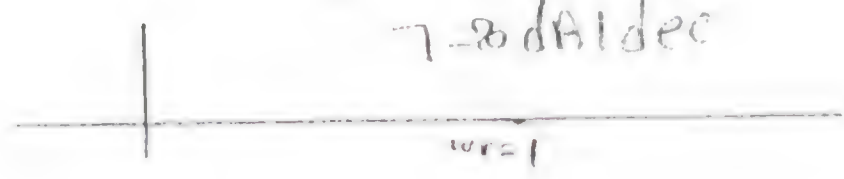

$$\rightarrow \text{put } z = \frac{1+r}{1-r}$$

$$G H(r) = 0.5 \frac{(\frac{1+r}{1-r} + 0.76)}{(\frac{1+r}{1-r} - 1)(\frac{1+r}{1-r} - 0.45)} = 0.5 \frac{(1+r + 0.76 - 0.76r)(1-r)}{(1+r-1+r)(1+r-0.45+0.45r)}$$

$$= 0.5 \frac{(1-r)(1.76 + 0.24r)}{2r(0.55 + 1.45r)} = \frac{0.5 \times 1.76}{2 \times 0.55} \frac{(1-r)(1 + \frac{0.24}{1.76}r)}{(1 + \frac{1.45}{0.55}r)r}$$

$$= 0.8 \frac{(1-r)(1 + \frac{r}{7.33})}{r(1 + \frac{r}{0.379})}$$

$$G H(j\omega) = \frac{0.8(1-j\omega r)(1 + \frac{j\omega r}{7.33})}{j\omega r(1 + \frac{j\omega r}{0.379})}$$

Term	$Q(\omega r)$	
$0.8$	$0$	$\} 20 \log 0.8$
$1 - j\omega r$	$-\tan^{-1} \frac{\omega r}{1}$	
$1 + j \frac{\omega r}{7.33}$	$\tan^{-1} \frac{\omega r}{7.33}$	
$\frac{1}{j\omega r}$	$-90$	
$1 / (1 + \frac{j\omega r}{0.379})$	$-\tan^{-1} \frac{\omega r}{0.379}$	

$$Q(\omega r) = -\tan^{-1} \omega r - \tan^{-1} \frac{\omega r}{0.379} + \tan^{-1} \frac{\omega r}{7.33} - 90$$

$\omega r$	0	1	0.379	7.33	0.5	5	$\infty$
$Q$	-90	-196.625	-152.796	-214.271	-165.5	-220.05	-180



$z^2 Y(z) + 6z Y(z) + 5 Y(z) = 2 R(z)$

sheet (6)(3)

2) for the following system → Controllable, observable, Draw State Diagram

a)  $Y(K+2) + 6Y(K+1) + 5Y(K) = 2R(K)$

$z \cdot T \quad \downarrow$

$$z^2 Y(z) + 6z Y(z) + 5 Y(z) = 2 R(z)$$

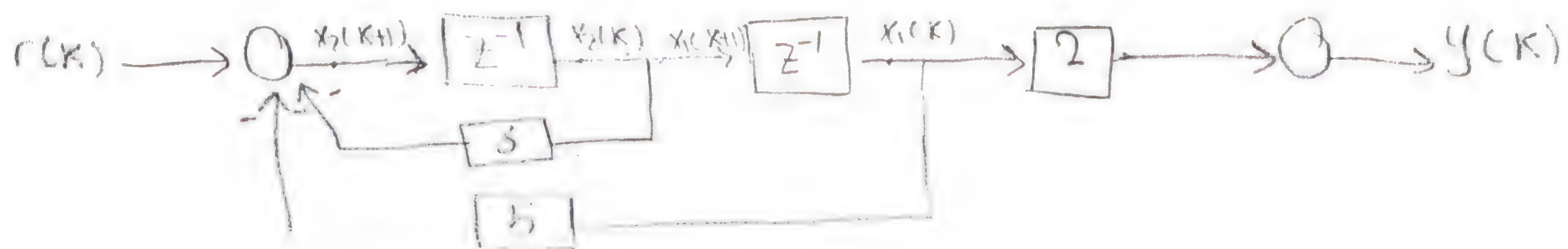
$$Y(z) [z^2 + 6z + 5] = 2 R(z)$$

$$\frac{Y(z)}{R(z)} = \frac{2}{z^2 + 6z + 5}$$

1) controllable form

$$X(K+1) = \begin{pmatrix} 0 & 1 \\ -5 & -6 \end{pmatrix} X(K) + \begin{pmatrix} 0 \\ 1 \end{pmatrix} R(K)$$

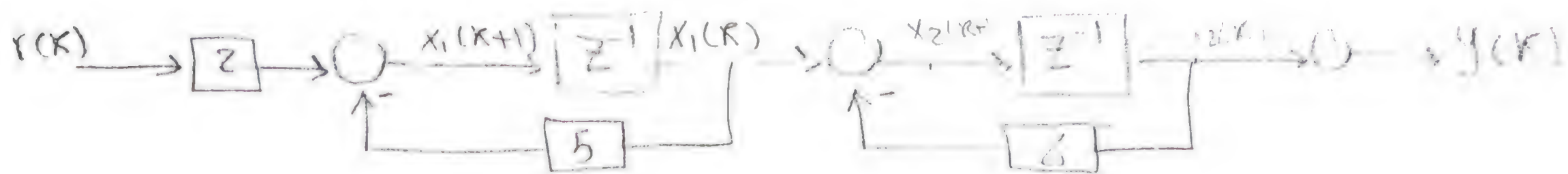
$$Y(K) = (2 \quad 0) X(K)$$



2) observable form

$$X(K+1) = \begin{pmatrix} 0 & -5 \\ 1 & -6 \end{pmatrix} X(K) + \begin{pmatrix} 2 \\ 0 \end{pmatrix} R(K)$$

$$Y(K) = (0 \quad 1) X(K)$$





$$(b) y(k+2) + 6y(k+1) + 5y(k) = 3r(k+2) + r(k+1) + 2r(k)$$

$$z^2 Y(z) + 6z Y(z) + 5Y(z) = 3z^2 R(z) + z R(z) + 2R(z)$$

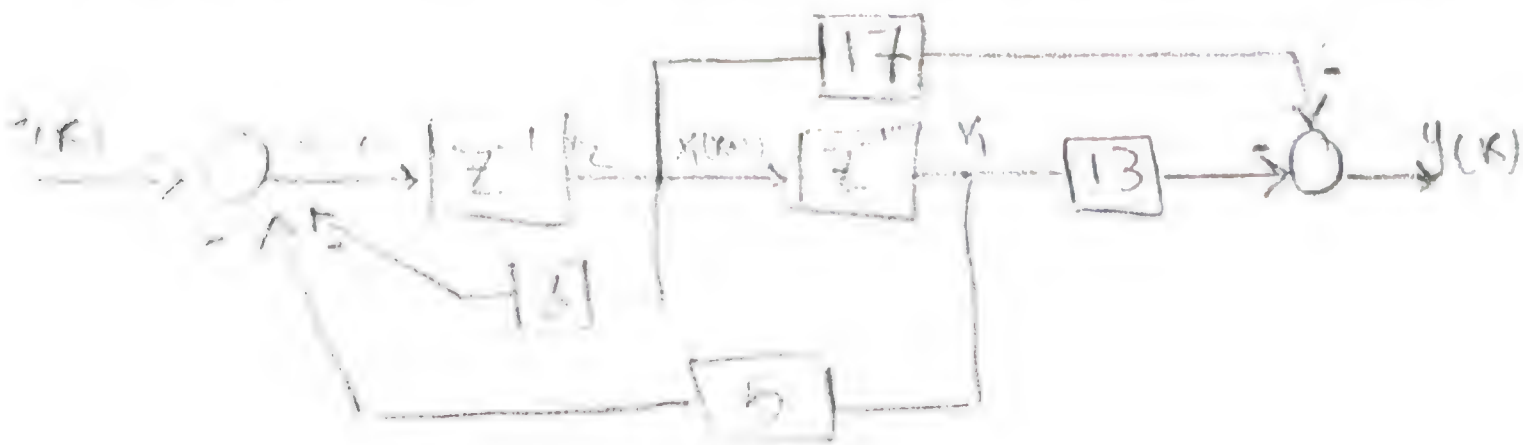
$$Y(z) [z^2 + 6z + 5] = R(z) [3z^2 + z + 2]$$

$$\frac{Y(z)}{R(z)} = \frac{3z^2 + z + 2}{z^2 + 6z + 5}$$

controllable form

$$X(k+1) = \begin{pmatrix} 0 & 1 \\ -5 & -6 \end{pmatrix} X(k) + \begin{pmatrix} 0 \\ 1 \end{pmatrix} r(k)$$

$$y(k) = (-13 \quad -17) X(k) + 3 r(k)$$



$$\begin{array}{r} 3 \quad 1 \quad 2 \\ \uparrow \quad \uparrow \quad \uparrow \\ B_0 z^2 + B_1 z + B_2 \\ \hline z^2 + 6z + 5 \end{array}$$

$$y(k) = A_1 x_1(k) + A_2 x_2(k) + B_0 r(k)$$

$$A_1 = B_2 - B_0 d_2$$

$$A_2 = B_1 - B_0 d_1$$

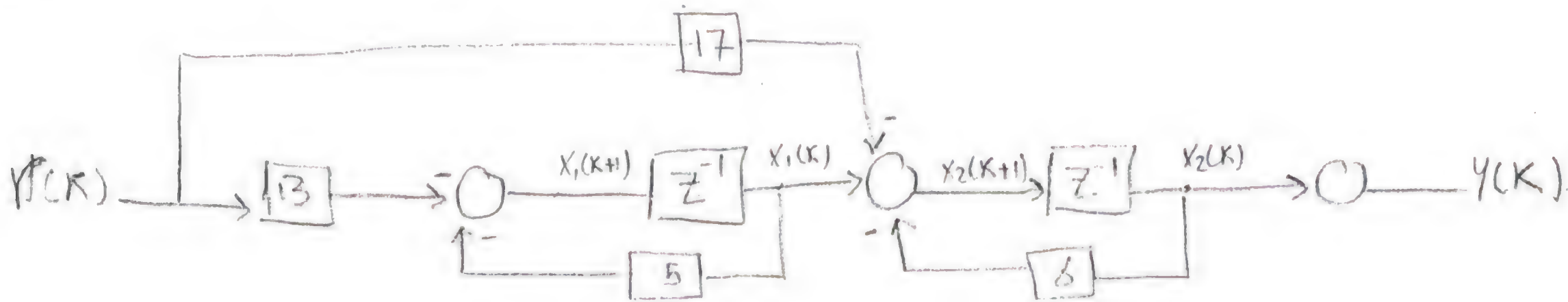
$$A_1 = 2 - 3 \cdot 5 = 2 - 15 = -13$$

$$A_2 = 1 - 3 \cdot 6 = 1 - 18 = -17$$

2) observable form

$$X(k+1) = \begin{pmatrix} 0 & -5 \\ 1 & -6 \end{pmatrix} X(k) + \begin{pmatrix} -13 \\ -17 \end{pmatrix} r(k)$$

$$y(k) = (0 \quad 1) X(k) + 3 r(k)$$





sheet (6)(4)

$$(c) \frac{Y(z)}{R(z)} = \frac{z^2 + 2}{z^3 + 3z^2 - 4z + 2}$$

$$\frac{Y(z)}{V(z)} \cdot \frac{V(z)}{R(z)} = \frac{z^2 + 2}{z^3 + 3z^2 - 4z + 2}$$

1) controllable form

$$\rightarrow Y(z) = z^2 V(z) + 2V(z)$$

$$y(k) = v(k+2) + 2v(k)$$

$$y(k) = x_3(k) + 2x_1(k)$$

$$\rightarrow R(z) = z^3 V(z) + 3z^2 V(z) - 4z V(z) + 2V(z)$$

$$R(k) = v(k+3) + 3v(k+2) - 4v(k+1) + 2v(k)$$

$$R(k) = x_3(k+1) + 3x_3(k) - 4x_2(k) + 2x_1(k)$$

Assum

$$v(k) = x_1(k)$$

$$v(k+1) = x_2(k)$$

$$v(k+2) = x_3(k)$$

$$v(k+3) = x_4(k)$$

$$x_1(k+1) = x_2(k)$$

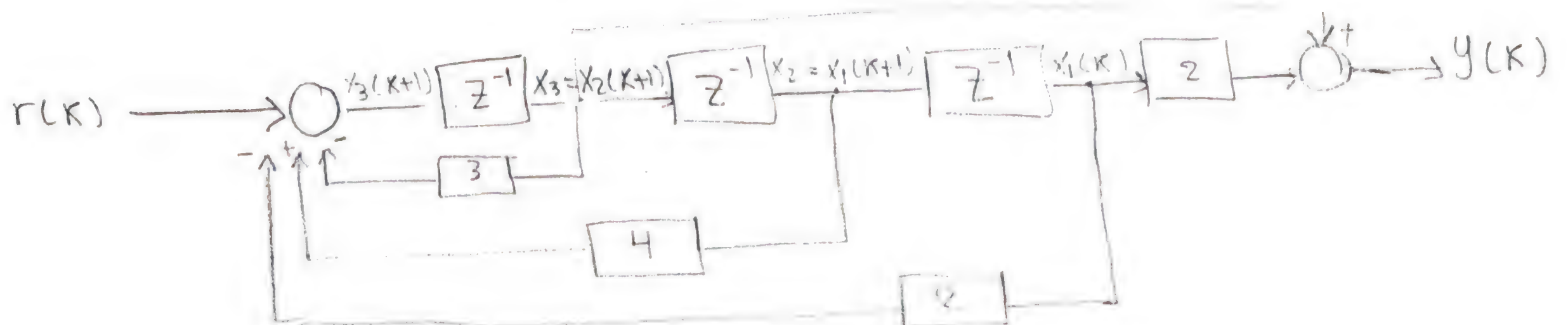
$$x_2(k+1) = x_3(k)$$

$$x_3(k+1) = R(k) - 3x_3(k) + 4x_2(k) - 2x_1(k)$$

$$x_4(k+1) =$$

$$x(k+1) = \begin{pmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -2 & 4 & -3 \end{pmatrix} \begin{pmatrix} x_1(k) \\ x_2(k) \\ x_3(k) \end{pmatrix} + \begin{pmatrix} 0 \\ 0 \\ 1 \end{pmatrix} r(k)$$

$$y(k) = \begin{pmatrix} 2 & 0 & 1 \end{pmatrix} \begin{pmatrix} x_1(k) \\ x_2(k) \\ x_3(k) \end{pmatrix} + \begin{pmatrix} 0 \\ 0 \\ 0 \end{pmatrix} r(k)$$





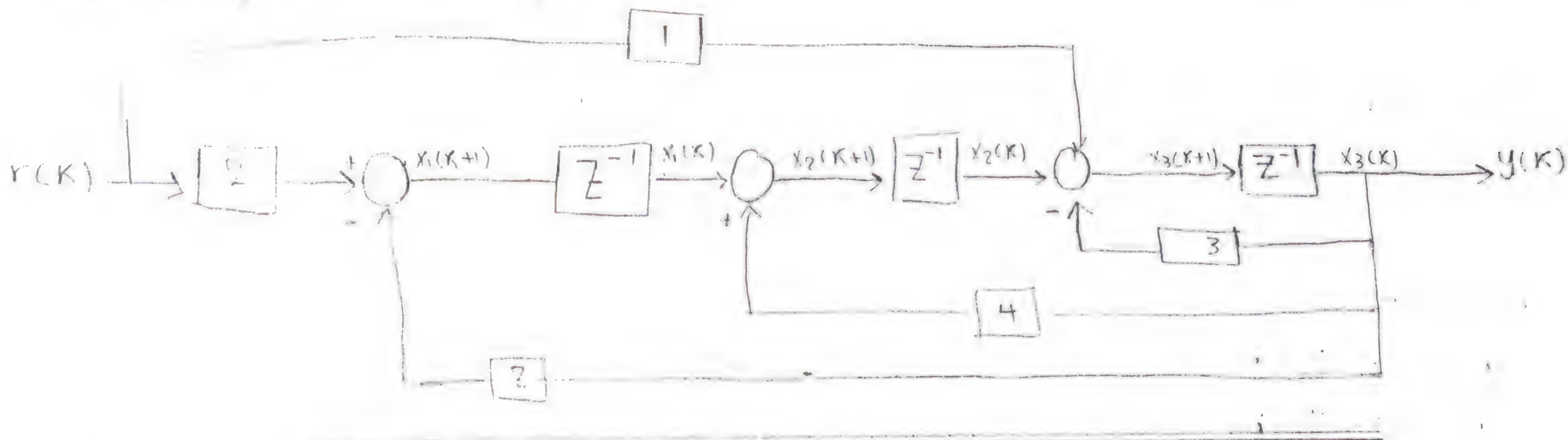
$$\begin{pmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -2 & 4 & -3 \end{pmatrix}$$

$$(2 \ 0 \ 1)$$

2) observable form

$$x(k+1) = \begin{pmatrix} 0 & 0 & -2 \\ 1 & 0 & 4 \\ 0 & 1 & -3 \end{pmatrix} \begin{pmatrix} x_1(k) \\ x_2(k) \\ x_3(k) \end{pmatrix} + \begin{pmatrix} 2 \\ 0 \\ 1 \end{pmatrix} r(k)$$

$$y(k) = (0 \ 0 \ 1) x(k)$$

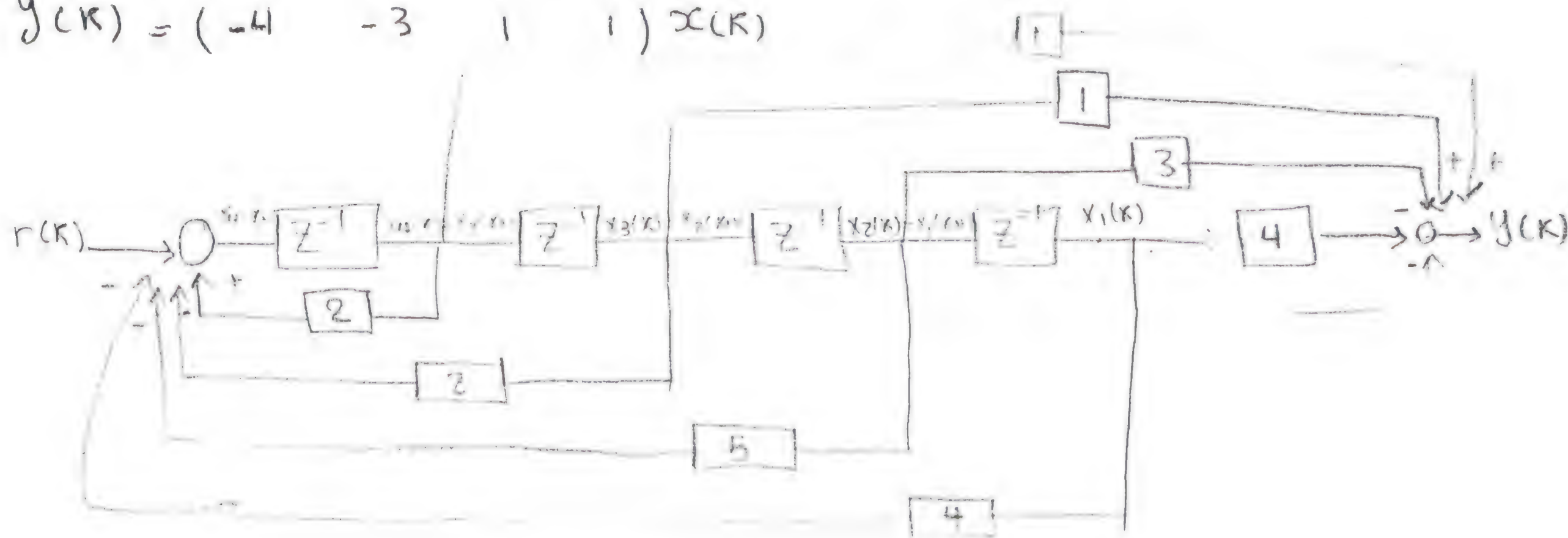


$$(d) \frac{Y(z)}{R(z)} = \frac{z^3 + z^2 - 3z - 4}{z^4 - 2z^3 + 2z^2 + 5z + 4}$$

1- Controllable form

$$x(k+1) = \begin{pmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ -4 & -5 & -2 & 2 \end{pmatrix} \begin{pmatrix} x_1(k) \\ x_2(k) \\ x_3(k) \\ x_4(k) \end{pmatrix} + \begin{pmatrix} 0 \\ 0 \\ 0 \\ 1 \end{pmatrix} r(k)$$

$$y(k) = (-4 \ -3 \ 1 \ 1) x(k)$$



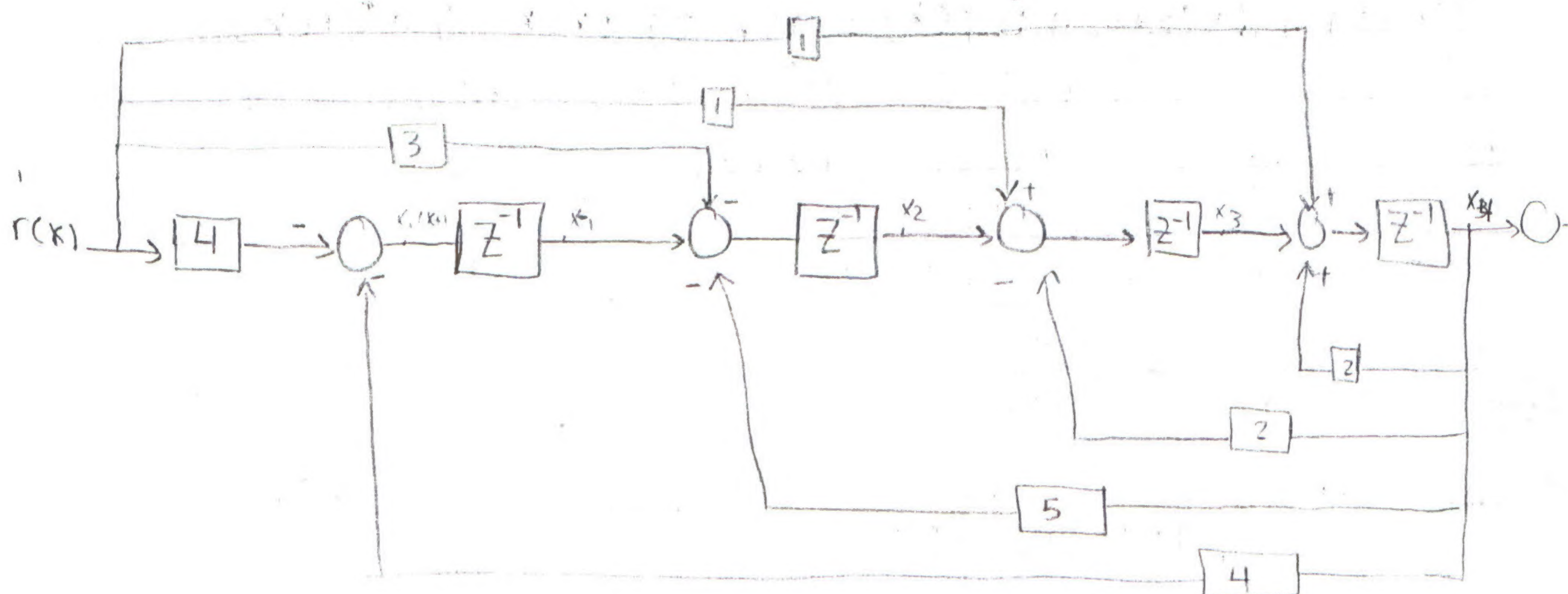


$$\begin{matrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ \text{sheet}(b) & (s) & & 1 \end{matrix}$$

2. observable form

$$x(k+1) = \begin{pmatrix} 0 & 0 & 0 & -4 \\ 1 & 0 & 0 & -5 \\ 0 & 1 & 0 & -2 \\ 0 & 0 & 1 & 2 \end{pmatrix} x(k) + \begin{pmatrix} -4 \\ -3 \\ 1 \\ 1 \end{pmatrix} r(k)$$

$$y(k) = (0 \ 0 \ 0 \ 1) x(k)$$



③ obtain the Diagonal State Space form & Draw the State Diagram

$$(a) y(k+2) + 6y(k+1) + 5y(k) = 2r(k)$$

$$z^2 y(z) + 6z y(z) + 5y(z) = 2R(z)$$

$$y(z) [z^2 + 6z + 5] = 2R(z)$$

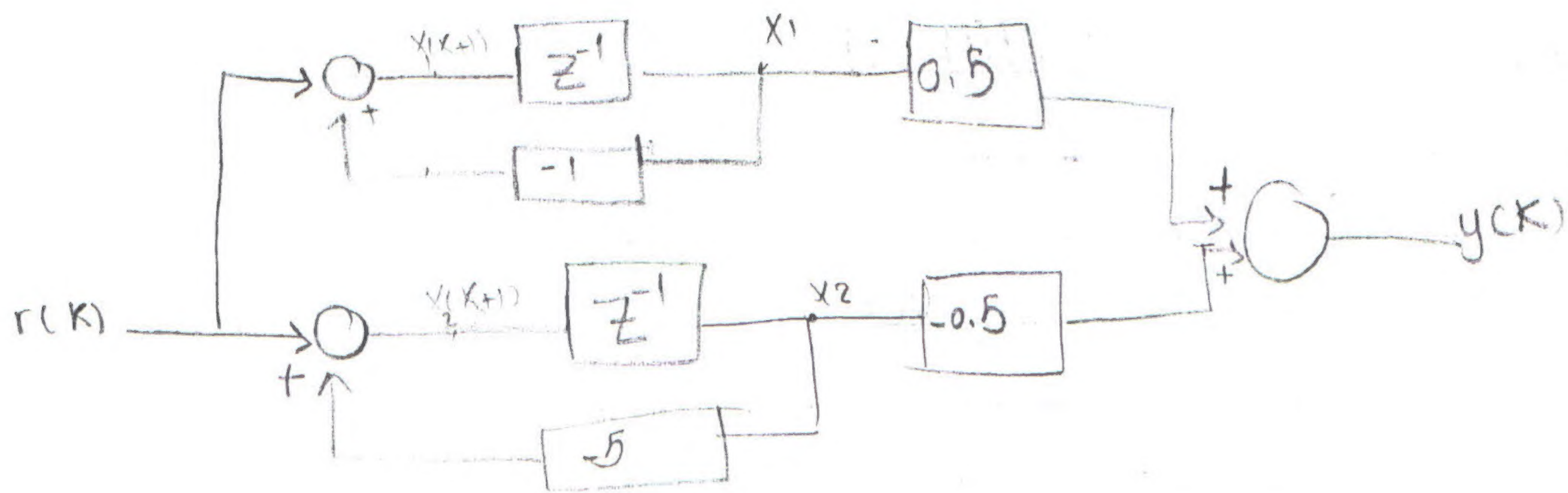
$$\frac{y(z)}{R(z)} = \frac{2}{z^2 + 6z + 5} = 2 \left[ \frac{A}{(z+1)} + \frac{B}{(z+5)} \right] = 2 \left[ \frac{1}{(z+1)(z+5)} \right]$$

$$= 2 \left[ \frac{0.25}{z+1} - \frac{0.25}{z+5} \right] = \frac{0.5}{z+1} - \frac{0.5}{z+5}$$

$$x(k+1) = \begin{pmatrix} -1 & 0 \\ 0 & -5 \end{pmatrix} \begin{pmatrix} x_1(k) \\ x_2(k) \end{pmatrix} + \begin{pmatrix} 1 \\ 1 \end{pmatrix} r(k)$$

$$y(k) = (0.5 \ -0.5) x(k)$$





$$(b) \quad y(k+2) + 6y(k+1) + 5y(k) = 3r(k+2) + r(k+1) + 2r(k)$$

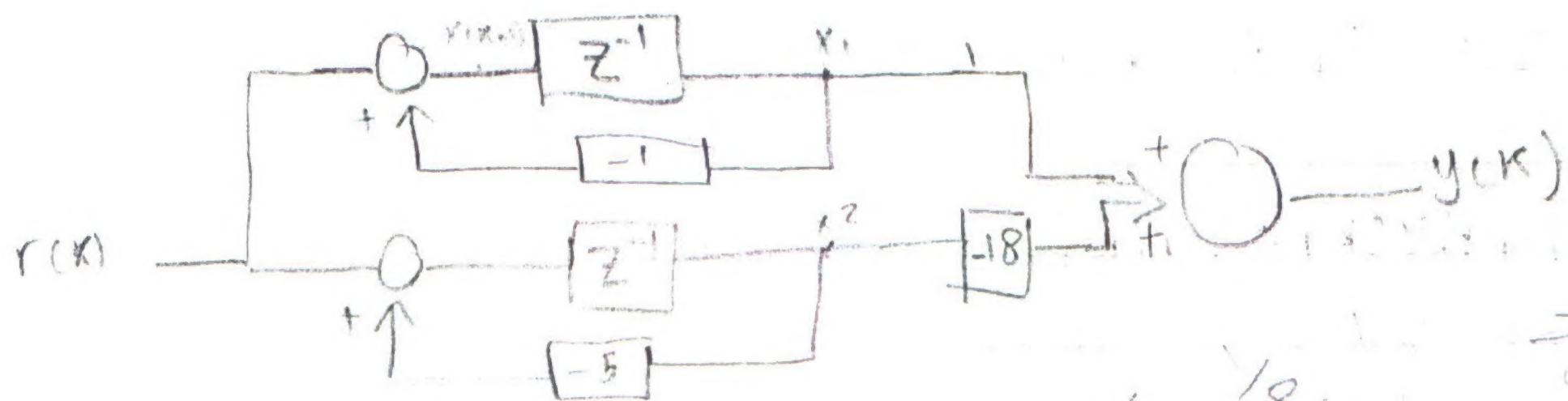
$$z^2 y(z) + 6zy(z) + 5y(z) = 3z^2 R(z) + zR(z) + 2R(z)$$

$$y(z) [z^2 + 6z + 5] = R(z) [3z^2 + z + 2]$$

$$\frac{y(z)}{R(z)} = \frac{3z^2 + z + 2}{z^2 + 6z + 5} = \frac{3z^2 + z + 2}{(z+1)(z+5)}$$

$$\frac{y(z)}{R(z)} = \frac{1}{z+1} + \frac{-18}{z+5}$$

$$x(k+1) = \begin{pmatrix} -1 & 0 \\ 0 & -5 \end{pmatrix} x(k) + \begin{pmatrix} 1 \\ 1 \end{pmatrix} r(k) \quad y(k) = (1 \quad -18) x(k)$$



$$(c) \quad \frac{y(z)}{R(z)} = \frac{z - 0.5}{z(z+0.5)(z+0.25)} = \frac{A}{z} + \frac{B}{z+0.5} + \frac{C}{z+0.25}$$

$$x(k+1) = \begin{pmatrix} 0 & 0 & 0 \\ 0 & -0.5 & 0 \\ 0 & 0 & -0.25 \end{pmatrix} x(k) + \begin{pmatrix} 1 \\ 1 \\ 1 \end{pmatrix} r(k)$$

$$y(k) = (0.66 \quad 0.125 \quad 12) x(k)$$



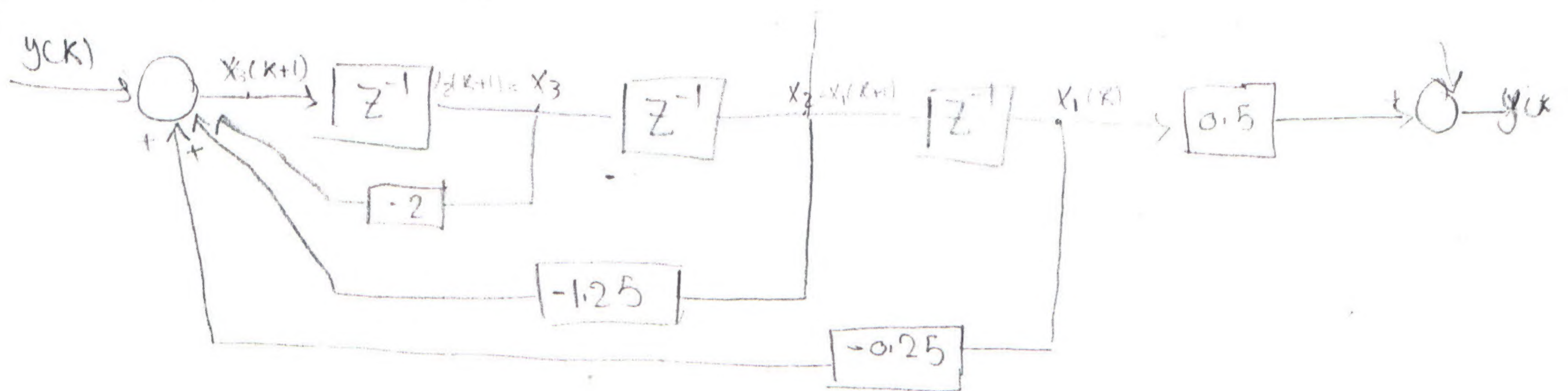
Discrete system  
1 sec

$$\frac{Y(z)}{R(z)} = \frac{z - 0.5}{(z + 0.5)^2 (z + 1)} = \frac{z - 0.5}{z^3 + 2z^2 + 1.25z + 0.25}$$

≡ controllable

$$X(k+1) = \begin{pmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -0.25 & -1.25 & -2 \end{pmatrix} X(k) + \begin{pmatrix} 0 \\ 0 \\ 1 \end{pmatrix} r(k)$$

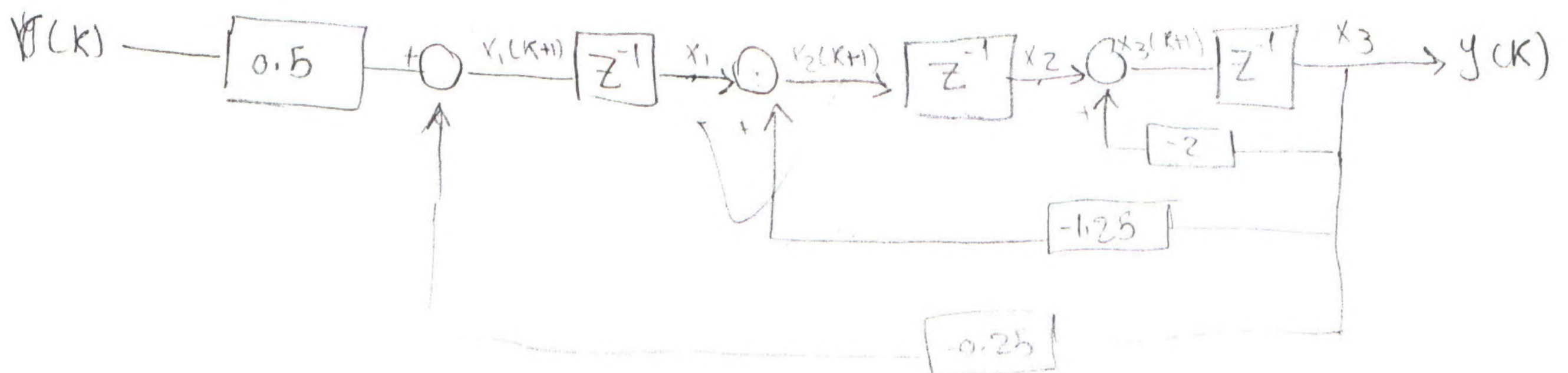
$$y(k) = (0.5 \quad 1 \quad 0) X(k)$$



≡ observable

$$X(k+1) = \begin{pmatrix} 0 & 0 & -0.25 \\ 1 & 0 & -1.25 \\ 0 & 1 & -2 \end{pmatrix} X(k) + \begin{pmatrix} 0.5 \\ 1 \\ 0 \end{pmatrix} r(k)$$

$$y(k) = (0 \quad 0 \quad 1) X(k)$$



≡ Diagonal =  $\frac{A_{11}}{(z + 0.5)^2} + \frac{B}{z + 0.5} + \frac{C}{z + 1}$

$$X(k+1) = \begin{pmatrix} 1 & 0 & 0 \\ 0 & -0.5 & 0 \\ 0 & 0 & -1 \end{pmatrix} X(k) + \begin{pmatrix} 0 \\ 1 \\ 1 \end{pmatrix} r(k)$$



$$y(k) = \begin{pmatrix} A & B & C \end{pmatrix} x(k)$$

$$\begin{pmatrix} -6 & 6 & -2 \end{pmatrix}$$

